

**Modeling sound scattering using a combination of the edge source integral equation
and the boundary element method**

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1 A hybrid method for sound scattering calculations is presented in this paper. The
2 boundary element method (BEM) is combined with a recently developed edge source
3 integral equation (ESIE) [J. Acoust. Soc. Am. 133, pp. 3681-3691, 2013]. Although
4 the ESIE provides accurate results for convex, rigid polyhedra, it has several numer-
5 ical challenges, one of which applies to certain radiation directions. The proposed
6 method, denoted ESIEBEM, overcomes this problem with certain radiation directions
7 by applying a similar approach as BEM. First, the sound pressure is calculated on
8 the surface of the scattering object using the ESIE, then second, the scattered sound
9 is obtained at the receiver point using the Kirchhoff-Helmholtz boundary integral
10 equation, as BEM does. The three methods have been compared for the scattering
11 by a rigid cube. Based on results from several discretizations, ESIE and ESIEBEM
12 results are typically (90% quartile) within $3 - 4 \cdot 10^{-4}$ for a kL -value of 1.83 and
13 $2 \cdot 10^{-3}$ for $kL = 9.15$, L being the cube length, of reference results computed with
14 the BEM. The computational cost of ESIEBEM appears to be lower than BEM.

15 I. INTRODUCTION

16 Accurate and fast numerical modelling of sound propagation and scattering is of great
17 interest nowadays. A wide range of problems ranging from environmental acoustic prob-
18 lems to musical instrument synthesis require the modeling of large scale 3D computational
19 domains. The computational complexity of standard solution methods such as the finite
20 element method (FEM), boundary element method (BEM) and finite differences in the time
21 domain (FDTD), scale poorly with the problem size. Therefore, alternatives to these well
22 known methods are needed. The fast multipole boundary element method (FMBEM) is
23 an alternative method to accelerate the calculations of the boundary element method¹ by
24 clustering boundary elements and using multipole expansions to evaluate the interactions
25 among clusters.

26 A recent edge source integral equation method (ESIE) presented by Asheim and Svensson²
27 has shown to be more efficient computationally than the numerical methods mentioned
28 above for convex, rigid scattering objects. Instead of using a mesh for the whole body
29 surface (BEM), and possibly also the air surrounding the object (FEM, FDTD), only a
30 discretization of the object edges is needed to compute so-called edge source amplitudes.
31 In a subsequent stage, these edge source amplitudes can be used for computing the sound
32 pressure in any external receiver positions, similar to the BEM, where the surface sound field
33 can be used in a similar way. It has not been possible to prove that the ESIE method fulfills
34 the governing Helmholtz equation, only that the results are remarkably accurate for rigid,

35 convex scattering bodies. Although this method is very attractive computationally, and in
36 terms of accuracy in general, erroneous results may arise for certain receiver positions³.

37 The goal of this paper is to propose a method that achieves fast calculation while main-
38 taining good accuracy. The method developed is a combination of the edge source integral
39 equation (ESIE) method and the boundary element method (BEM). Parts of this work were
40 presented in the POMA paper⁴ by the same authors. The method avoids the singularities for
41 certain receiver positions of the ESIE³, and it also avoids the well-known internal resonance
42 phenomenon of the BEM formulation commonly avoided by using the Burton and Miller’s
43 method⁵ or the CHIEF points technique⁶. The computational cost of this hybrid method
44 will be compared to the BEM and to the original ESIE.

45 This paper is organized as follows: in section II, the fundamental equations and theo-
46 retical backgrounds for the three methods, the boundary element method (BEM), the edge
47 diffraction-based ESIE method and the hybrid method are introduced. Section III summa-
48 rizes the implementation details of the methods as well as the description of the benchmark
49 case, and the results obtained are presented and discussed in section IV. Finally, section V
50 collects the conclusions of the paper.

51 II. THEORY

52 There are two large families of methods for solving acoustic scattering problems: those
53 commonly called wave-based techniques derived from the wave equation (BEM, FDTD,
54 FEM, FMBEM) and those referred to as geometrical-acoustics techniques which are high
55 frequency asymptotic solutions (the image source method, ray tracing). The latter can also

56 have diffraction-based extensions such as the geometrical theory of diffraction (GTD) pre-
57 sented by Keller⁷, the Uniform Theory of Diffraction (UTD) by Kouyoumjian and Pathak⁸,
58 or the recent edge source integral equation method (ESIE) by Asheim and Svensson² men-
59 tioned above. Two of these methods will be presented more in detail below: the wave-based
60 boundary element method, and the edge source integral equation method. Their respective
61 advantages and drawbacks will be identified, and a new hybrid method will be presented
62 that exploits the advantages of both techniques and avoids some of their limitations when
63 combined to solve scattering problems.

64 **A. The boundary element method (BEM)**

65 The acoustic boundary element method is a well-established method in acoustics, espe-
66 cially suitable for infinite domains (outdoor/free-field environment). The BEM formulation
67 is based on the Helmholtz integral equation which relates the sound pressure $p(P)$ at any
68 point P to the sound pressure $p(Q)$ and the normal velocity $v_n(Q)$ at positions Q on the
69 surface D of a scattering body. In this paper, only rigid scattering polyhedra will be stud-
70 ied and so the Neumann condition $v_n(Q) = 0$ holds for any point $Q \in D$. Therefore, the
71 Helmholtz integral equation for these cases can be expressed without the monopole term
72 and it can be written as follows⁹

$$C(P)p(P) = 4\pi p^I(P) + \int_D \frac{\partial G(P, Q)}{\partial \mathbf{n}} p(Q) dS, \quad (1)$$

73 where $G(P, Q) = \frac{e^{-jkR}}{R}$ is the free-field Green's function in 3D between two points P and
74 Q , R is the Euclidean distance between P and Q , p^I is the incident sound pressure, $C(P)$

75 is the solid angle corresponding to 0 , 2π or 4π if P is inside, on the surface, or outside of
76 the object, respectively, k is the wavenumber, and \mathbf{n} is the normal vector to the surface at
77 Q pointing away from the body. A time-harmonic factor $e^{j\omega t}$ has been omitted in Eq. (1)
78 and throughout the paper.

79 The sound pressure at any point P expressed in Eq. (1) is interpreted here as a sum of
80 contributions from free-field radiating dipoles aligned with the surface normal vector. The
81 strength of the dipole is given by the surface sound pressure and once this sound pressure
82 on the surface is known, then the sound pressure at any point P can be obtained.

83 The boundary element method calculates the sound pressure in a field point P in two
84 steps. In the first step, the solution on the surface of the scattering object is obtained by
85 placing P on the surface of the scattering object, and Eq. (1) becomes an integral equation.
86 In the second step, the so-called "propagation" step, the sound pressure in the sound field
87 is calculated by letting P be an external point, often termed as a "field point".

88 Different discretization methods are commonly used for the first step, such as the projec-
89 tion methods (Galerkin, collocation) or Nyström methods, which turn Eq. (1) into various
90 forms of linear systems of equations. The number of degrees of freedom in that system of
91 equations is directly related to both the accuracy of the solution and the computational
92 cost.

93 The BEM, when it is applied to exterior scattering problems, has a well-known problem
94 as mentioned in the introduction. The matrix equation to solve becomes ill-conditioned at
95 the natural frequencies of the corresponding interior problem, but two different solutions
96 have been presented in the literature for this problem. A first technique is using so-called

97 CHIEF points, or internal control points, where the sound field is enforced to be zero⁶.
98 A second one is the so-called Burton-Miller technique⁵ which uses a linear combination of
99 the Kirchhoff-Helmholtz integral equation and its normal derivative. Marburg and Amini¹⁰
100 show that the Burton and Miller method is a more robust method compared to the CHIEF
101 method as its solution is unique for exterior acoustic problems at all frequencies¹⁰.

102 Another challenge of the BEM is the singularity in the integral kernels, which becomes
103 prominent for thin bodies and narrow gaps¹¹ as well as when the field point is located near
104 the boundary of the scattering object. There have been different approaches to overcome
105 that difficulty making use, for instance, of singular numerical integration as suggested by
106 Cutanda et al.¹², splitting the integral using analytical removal of the singularity^{11,13}, or
107 using a polar coordinates transformation as presented by T. Terai¹⁴. The details of these
108 techniques will not be discussed in this paper and the interested reader is referred to the
109 cited work for more details.

110 **B. Edge source integral equation method (ESIE)**

111 The edge-diffraction based method used in this paper is the edge source integral equation
112 (ESIE) method suggested recently by Asheim and Svensson². This method, which was
113 shown to be accurate and efficient for rigid convex scattering objects, decomposes the total
114 acoustic field into three different components:

$$p_{tot}(P) = p_{GA}(P) + p_{D1}(P) + p_{HOD}(P), \quad (2)$$

115 where $p_{GA}(P)$ is the geometrical acoustics component, and $p_{D1}(P)$ and $p_{HOD}(P)$ correspond
 116 to the first- and higher-order diffraction components respectively. The geometrical acoustics
 117 term, p_{GA} , represents the direct and reflected sound considering the visibility between source
 118 and receiver, and it can easily be obtained by the commonly used image source (IS) method.

119 The first-order diffraction is based on a representation of the diffracted field for a single
 120 wedge¹⁵, which is a reformulation of the analytical solution for infinite edges by Bowman
 121 and Senior¹⁶ that can be applied to finite edges.

122 This first-order diffraction term in Eq. (2) at a receiver's position P , $p_{D1}(P)$, can thus
 123 be computed as an explicit line-integral equation over the set of edges Γ of the scattering
 124 object as

$$p_{D1}(P) = -\frac{1}{4\pi}q_S\nu_z \times \int_{\Gamma} V_{P,z}V_{z,S} \frac{e^{-jkr_{P,z}}}{r_{P,z}} \frac{e^{-jkr_{z,S}}}{r_{z,S}} \beta(P, z, S) dz, \quad (3)$$

125 where q_S is the source strength of the sound source, S , defined such that $q_S = \rho_0 A/4\pi$,
 126 where ρ_0 is the density of the medium at rest, and A is the volume velocity amplitude of the
 127 monopole sound source. In Eq. (3) ν_z is the so-called wedge index, $V_{a,b}$ is a point-to-point
 128 visibility term being one when a is visible from b and zero otherwise. The term $\beta(P, z, S)$ is
 129 a function which depends only on the wedge angle and on the angles of the sound source, S ,
 130 and receiver, P , defined relative to the tangent at the edge point z , and thus β is interpreted
 131 as a directivity function of a virtual/secondary edge source at point z ¹⁵. The integral in

Eq. (3) can be computed by standard quadrature methods or, as suggested by Asheim and Svensson, by employing the efficient and accurate numerical method of steepest descent¹⁷.

The higher-order diffraction term, p_{HOD} , is obtained by the introduction of explicit edge source strengths for the secondary sources along the edges referred to above, for first-order diffraction. The computation process is similar to the BEM in that first, the strengths of these secondary edge sources are calculated by solving an integral equation and secondly, the diffracted sound pressure, p_{HOD} , is obtained via a propagation integral from the edge sources to the receiver. Below is a brief description of the formulation to obtain p_{HOD} .

Let $q(z_1, z_2)$ be defined as the equivalent source strength at an edge point z_2 radiating in the direction of another edge point z_1 . As shown by Asheim and Svensson², $q(z_1, z_2)$ needs to satisfy the following integral equation

$$q(z_1, z_2) = q_0(z_1, z_2) - \frac{1}{8\pi} \int_{\Gamma} q(z_2, z) \cdot \frac{e^{-jkr_{z_2, z}}}{r_{z_2, z}} \times \nu_{z_2} V_{z_1, z_2} V_{z_2, z} \beta(z_1, z_2, z) ds_z. \quad (4)$$

The term $q_0(z_1, z_2)$ corresponds to the equivalent source strength at z_2 due to the field diffracted at z_2 , coming from the source S , in the direction of z_1 which is expressed as follows

$$q_0(z_1, z_2) = -\frac{1}{8\pi} \nu_2 V_{z_1, z_2} V_{z_2, S} \frac{e^{-jkr_{z_2, S}}}{r_{z_2, S}} \beta(z_1, z_2, S). \quad (5)$$

Once the edge source strengths are obtained, by solving the integral equation (4), the term $p_{HOD}(P)$ can be computed by a double integral, each integration taken along the set of all edges, expressed as follows

$$p_{HOD}(P) = -\frac{1}{8\pi} \int_{\Gamma} \int_{\Gamma} q(z_1, z_2) \nu_{z_1} V_{P, z_1} V_{z_1, z_2} \times \frac{e^{-jkr_{P, z_1}}}{r_{P, z_1}} \frac{e^{-jkr_{z_1, z_2}}}{r_{z_1, z_2}} \beta(P, z_1, z_2) ds_{z_2} ds_{z_1}, \quad (6)$$

146 where the spherical radiation factors and the same directivity function β as used in Eq.(3)
 147 have been considered.

148 Note that the computation of the edge source strengths $q(z_1, z_2)$ in Eq.(4) is independent
 149 of the receiver's position in the same way as the first step of the BEM when computing the
 150 sound field at the surface of the scattering object. The attractiveness of the ESIE, though, is
 151 that there is only the need to discretize the edges of the object instead of the entire surface,
 152 which reduces considerably the computational cost. A further advantage of the ESIE versus
 153 the BEM is that the ESIE does not have any problem with the internal fictive resonances
 154 that the BEM suffers from. On the other hand, the intermediate quantities are more directly
 155 useful for the BEM than for the ESIE: the surface sound pressure might be exactly what is
 156 sought in some applications, whereas those edge source amplitudes are apparently not useful
 157 for anything by themselves.

158 The ESIE gives very accurate results for rigid convex scattering objects (so, with no
 159 indents) and even gives accurate results in the low-frequency limit². As mentioned in the
 160 Introduction, it has not been shown that the ESIE method should give an exact solution to
 161 the Helmholtz equation. Very accurate results have been demonstrated nevertheless, and
 162 it is not clear how this conundrum can be tackled. But, for certain receiver positions, the
 163 convergence to the accurate solution for the propagation step is very slow due to that the

164 directivity function β in Eq.(6) has some singularities that depend on the positions of the
 165 receiver P relative to the edge sources z_1 and z_2 . Two edges define a virtual plane, exterior
 166 to the scatterer, and the directivity function β makes a jump as a receiver position crosses
 167 that virtual plane³. Associated with that jump is a very slow convergence.

168 Since the β -function is also used in the expression for the term q_0 , in Eq. (5), there will
 169 be inaccurate results associated with external source positions that are very near to any of
 170 the planes that are formed by pairs of the scattering body's edges. Finally, the presence of
 171 the β -function in the integral equation operator in Eq. (4) leads to slow convergence when
 172 smooth scattering objects are represented by polyhedra. Interestingly, the ESIE formulation
 173 is numerically much more efficient for scattering bodies with edges than for smooth bodies.

174 C. Combining the ESIE and the BEM: ESIEBEM

175 It is possible to combine the two presented methods in a way that uses their respective
 176 strengths and overcomes some of their respective weaknesses. The hybrid method proposed
 177 here, called the ESIEBEM from now on, is based on using the ESIE, instead of the Helmholtz
 178 integral equation, for obtaining the sound pressure on the surface. At this calculation
 179 stage, the surface of the object is discretized by an element mesh and the sound pressure
 180 is computed at the element centers (collocation points). The expression of the directivity
 181 function β for this case, appearing in both Eq. (3) and Eq. (6), has been derived in the
 182 Appendix VII. Note that β would have no singularity related to the receivers positions on
 183 the surface, except for source positions close to a plane of the scattering object (as discussed
 184 in Section II B). There is also a $1/r$ -singularity, where r is the distance from the receiver

185 point to and edge point, which is the reason to use collocation points instead of the element
 186 nodes. However, the computation of the directivity function β will still have some challenges
 187 when two or more faces of the scattering polyhedron are close to co-planar.

188 The principles of the ESIE and ESIEBEM are illustrated in Fig. 1. The ESIEBEM
 189 involves three calculation steps. The first step is the same as for the ESIE: calculation of
 190 the edge source strengths, given the external source, using Eqs. (4) and (5). The second
 191 step is the edge source-based computation of the sound pressure in the element collocation
 192 points on the surface, using Eq. (2), (3), and (6) and the third step employs the propagation
 193 integral of the BEM, Eq. (1).

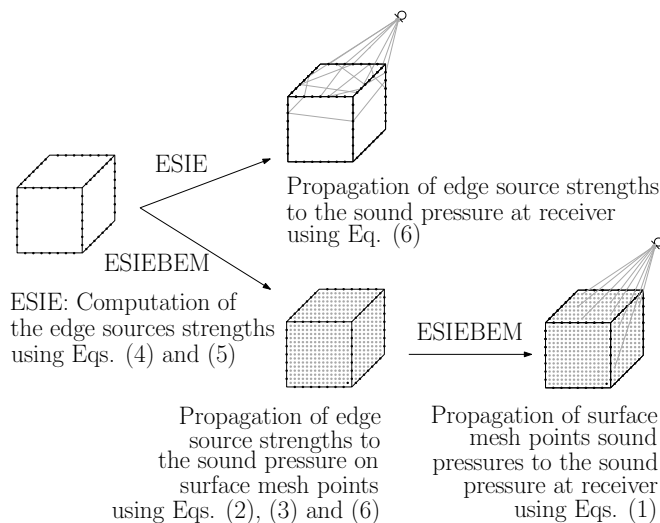


FIG. 1. Illustration of the ESIE and the ESIEBEM computation steps. The discretization of edges and surfaces as well as paths drawn are only a few, for illustration purposes.

194 The main advantage of using the ESIE in this first step of computing the sound pressure
 195 on the surface is that the ESIE is highly efficient for finding the sound pressure in receiver
 196 positions that are not challenging for the method. Therefore, the computation time might

197 be advantageous compared to the BEM. Moreover, the accuracy of the ESIEBEM seems to
 198 be guaranteed as long as the receiver points on the surface are not immediately at the edge³
 199 and the scatter has no co-planar faces.

200 III. IMPLEMENTATION

201 A. BEM - Direct collocation method

202 In this paper, the OpenBEM implementation developed by Juhl and Henriquez has been
 203 used for obtaining the BEM results. OpenBEM is a collection of open-source Matlab func-
 204 tions for solving acoustical problems in 2D, 3D or axi-symmetric settings¹⁸. Interested
 205 readers will find a detailed description of it in Ref.¹⁹ by Juhl and a shorter version of it in
 206 Ref.²⁰ by Henriquez and Juhl.

207 The implementation employs the direct collocation method to compute the sound pressure
 208 on the scattering body surface. The scatterer’s surface is discretized into a mesh of elements,
 209 triangular or quadrilateral, and the sound pressure is calculated at the nodes of this mesh.
 210 When the point P in Eq. (1) is placed at any node, the following matrix expression results

$$\mathbf{C}\mathbf{p} = \mathbf{A}\mathbf{p} + 4\pi\mathbf{p}^{\mathbf{I}}, \quad (7)$$

211 where the matrix \mathbf{A} contains integrals of the kernel functions defined in Eq. (1).

212 Since rigid scattering objects are considered in this paper, the Neumann boundary con-
 213 dition applies, i.e., $v_n(Q) = 0$, and the left-hand side term in the expression can be directly

214 subtracted from the diagonal of the first term on the right-hand side, and Eq. (7) is
215 simplified to

$$\mathbf{p} = \mathbf{D}^{-1}(-4\pi\mathbf{p}^{\mathbf{I}}), \quad (8)$$

216 where $\mathbf{D} = \mathbf{A} - \mathbf{C}$ is a full matrix. If the surface mesh used has N elements, then the
217 computational complexity to solve the scattering problem with the OpenBEM is of the
218 order of N^2 , for building up the needed matrices, with a subsequent step scaling as N^3
219 for inverting the matrix. The simple Gauss elimination is used here to solve Eq. (8),
220 although other more efficient iterative solvers could have been implemented, and reduced
221 the complexity by an order of magnitude. The minimum number of surface elements will
222 depend on the square of the maximum frequency studied, f_{\max} , so the total computation
223 time might scale as $T_{\text{comp}} \sim f_{\max}^4 - f_{\max}^6$, depending on which stage is computationally
224 dominating in a specific implementation.

225 The OpenBEM uses the common method with CHIEF points to avoid the problems at
226 certain fictive internal resonance frequencies⁶.

227 **B. ESIE - Matrix equation formulation**

228 The ESIE method used in this paper was implemented in Matlab by Svensson et al.²¹
229 as the "Edge diffraction toolbox" published under the terms of the GNU General Public
230 License and currently available on GitHub. The toolbox computes various combinations of
231 specular reflections and higher-order diffraction, in the frequency domain.

232 The ESIE is based on the solution of the integral equation in Eq. (4) where the unknown
 233 edge source strengths $q(z_1, z_2)$ need to be solved for each pair of edge points (z_1, z_2) . This is
 234 done using the straightforward Nyström, or quadrature, method, where the integral equation
 235 is discretized at positions z^i along all the straight edges of the polyhedron. The unknowns
 236 are the edge source strengths, defined for pairs of discrete points along the edge, $q(z_1^i, z_2^j)$,
 237 which refer to the edge source amplitude at edge point z_2^j , in the direction of edge point
 238 z_1^i . For each straight edge of the polyhedron, a Gauss-Legendre (G-L) quadrature scheme is
 239 employed. A certain G-L quadrature order is chosen for the longest edge, and proportionally
 240 lower orders are chosen for shorter edges. For a cube example, if each of the 12 edges are
 241 discretised according to a G-L quadrature order n_{gauss} , a total of $12 \cdot n_{gauss}$ edge points will
 242 consequently be generated. As discussed further below, each edge point of a cube can reach
 243 6 other edges, and thus $6 \cdot n_{gauss}$ other edge points, for a total of $72 \cdot n_{gauss}^2$ unknowns. By
 244 constructing a column-vector \mathbf{q} with all the terms $q(z_1^i, z_2^j)$, z_1^i and z_2^j being the discretization
 245 points of all edges, Eq. (4) can be rewritten as the matrix expression

$$\mathbf{q} = \mathbf{q}_0 + \mathbf{H}\mathbf{q}, \quad (9)$$

246 where the matrix \mathbf{H} contains sampled values of the kernel of the integrand operator in Eq.
 247 (4), including the weighting factors of the Gauss-Legendre quadrature rule. This matrix
 248 equation can be solved by inversion

$$\mathbf{q} = [\mathbf{I} - \mathbf{H}]^{-1}\mathbf{q}_0, \quad (10)$$

249 although the size of the \mathbf{H} -matrix often prohibits such a direct inversion. Fortunately, \mathbf{H}
 250 has a very sparse nature, which makes an iterative solution of Eq. (9) very efficient². Thus,
 251 iteration step n gives a term

$$\mathbf{q}_n = \mathbf{H}\mathbf{q}_{n-1}, \quad n \in \mathbb{N}, \quad (11)$$

252 where \mathbf{q}_0 is given by sampled values of Eq. (5), and all terms of the truncated iteration
 253 process are summed up to the final solution,

$$\mathbf{q}_{\text{final}} = \sum_{n=0}^{N_{\text{truncation}}} \mathbf{q}_n, \quad (12)$$

254 The final solution is then propagated to the receiver with Eq. (6), and the term p_{HOD}
 255 obtained will correspond exactly to the contribution of all orders of diffraction up to and
 256 including order $(N_{\text{truncation}} + 2)$. The matrix equation formulation for this propagation is

$$p = \mathbf{F}\mathbf{q}_{\text{final}}, \quad (13)$$

257 where p is the sound pressure amplitude in a single receiver point, and \mathbf{F} is a horizontal
 258 vector of samples of the integrand in Eq. (6), again with weighting factors according to the
 259 Gauss-Legendre quadrature.

260 The sparseness of the matrix \mathbf{H} is the reason for the efficiency of the ESIE, and it is
 261 explained in Appendix VIII. The minimum number of edge points/sources, N_{es} will scale as
 262 $N_{\text{es}} \sim f_{\text{max}}$, whereas the number of edge source amplitudes, N_q will scale as $N_q \sim N_{\text{es}}^2 \sim f_{\text{max}}^2$.
 263 The size of the \mathbf{H} -matrix is such that the number of non-zero terms is $\sim N_q^{3/2}$, so the iterative
 264 solution of the matrix equation will cost $T_{\text{comp}} \sim f_{\text{max}}^3$. This suggests that the ESIE could

265 indeed be more efficient than the BEM for computing the field at the surface of the scattering
266 body.

267 C. ESIEBEM

268 The hybrid method ESIEBEM suggested in this paper employs the discretization of the
269 scatterer's surface like the BEM does. Triangular elements are used in this paper rather than
270 quadrilateral elements but the hybrid method might be used with quadrilateral meshing as
271 well.

272 As mentioned earlier in the text, it is known that the ESIE converges very slowly for
273 receiver positions where the visibility factor suddenly changes from 0 to 1, along the zone
274 boundaries that extend away from the scattering polyhedron. Receiver positions on the
275 surface of the scattering bodies are, however, not exposed to this problem. On the other
276 hand, along the edges of the scattering polyhedron, there are numerical challenges for the
277 ESIE, caused by the $1/r$ -factor in the involved integrals. No scheme has been developed for
278 mitigating this singularity, and therefore the standard quadrature method that is employed
279 here becomes inefficient for receiver positions very close to the edge. However, if the surface
280 sound pressure is calculated at element center points (i.e. collocation points) rather than at
281 the nodes, the effect of this singularity is reduced. Improved schemes might be developed
282 that handle that singularity more efficiently.

283 Yet another singularity occurs at the corners where two or more edges meet. The Gauss-
284 Legendre quadrature approach does not use any quadrature points at the integration range
285 endpoints, that is, at those corners, and no problems have been encountered with the quadra-

286 ture used for the integral equation. A well-behaved polynomial convergence is demonstrated
287 in section IV B, as long as receiver points are not close to any zone boundaries. Also, in
288 Ref.², the case of a circular disc was studied in detail. For a symmetrical incident field, the
289 integral equation could be simplified to a one-dimensional one, which was evaluated with the
290 midpoint method, in order to avoid the same singularities at the endpoints of the integration
291 range. No problems with convergence were encountered there either.

292 The internal-resonance problem and the thin-scattering-body problem mentioned in sec-
293 tion II A that BEM encounters, do not apply to the ESIEBEM, but the near-singularity issue
294 for field points near the scattering bodies applies for both, BEM and ESIEBEM. OpenBEM
295 uses a refined quadrature scheme and the solution for the near-singular kernels could be
296 either to increase the mesh density near the close point or to increase the order of the
297 numerical integration? .

298 The test cases are calculated with different meshes when solving the problem with the
299 BEM and ESIEBEM, and with different numbers of edge discretization points when solving
300 the problem with the ESIE. The meshes have been created with the open-source GMSH
301 software²². All elements are considered to be triangular and the receiver points (i.e. the col-
302 location points of the elements) correspond to the center points of these triangular elements
303 in the first ESIEBEM calculation step. The ESIEBEM uses isoparametric elements with
304 constant shape functions and derivatives of these shape functions equal to zero, as employed
305 in the OpenBEM.

306 D. Test case

307 The scattering object studied here is one of the simplest to test the performance of the
308 novel hybrid method: a rigid cube as shown in Fig. 2. A cube of size $1 \times 1 \times 1 \text{ m}^3$ has been
309 modeled centered at the origin, assuming an incident field of a sound source located very far
310 away at a position $(x, y, z) = 10^6/\sqrt{3}(1, 1, 1) \text{ m}$, to emulate a plane wave impinging on the
311 scattering object. The results shown in this section are for the frequencies 100 and 500 Hz,
312 corresponding to $kL = 1.83$ and 9.15 , respectively, where L is the side length of the cube.

313 The receiver positions are in the plane $z = 0$ along a 1 m radius circumference with
314 a total of 629 receivers uniformly distributed, with a step of 0.01 radians, starting from 0.
315 Twelve different meshes were constructed for the BEM, with 408, 624, 744, 1096, 1480, 3124,
316 6260, 10204, 13084, 20456, 30940 and 50872 elements. With the ESIE method, edges were
317 discretized with 16, 24, 32, 40, 48, 56, 64, 80, and 96 edge points per edge, giving 192-1152
318 total edge points.

319 The calculations were carried out on a Macintosh HD with a processor of 2.7 GHz (Intel
320 Core i5) and 16 GB RAM and on a desktop computer with an operating system Windows
321 10 and a processor Intel(R) Xeon(R) 3.4 GHz and 8.0 GB RAM.

322 IV. RESULTS

323 In this section, results will first be presented with the finest BEM-mesh results viewed as
324 reference results. In the subsections following after, the application of linear extrapolation
325 will be explored to reach a higher accuracy for all the methods.

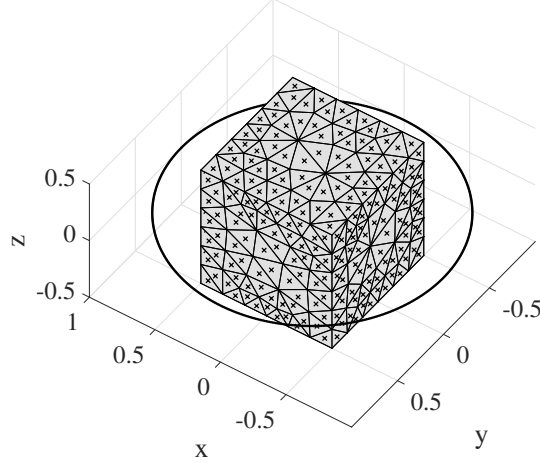


FIG. 2. Benchmark test case, a cube of $1 \times 1 \times 1 \text{ m}^3$ and 629 receivers located along a circumference of 1 m radius. Collocation points on the cube surface are depicted with black crosses. The same mesh with triangular elements is used both for the BEM and ESIEBEM, and the figure shows a coarse mesh for visualization purposes with 200 nodes and 396 collocation points.

326 **A. Overcoming problematic positions for the ESIE by using the ESIEBEM**

327 As explained in Section II B, the ESIE method has singularities for some receiver po-
 328 sitions. To demonstrate this effect, the sound pressure has been computed for the two
 329 frequencies, 100 Hz and 500 Hz, and all 629 receivers indicated in Fig. 2 using a BEM mesh
 330 with 50872 elements and collocation points, and an ESIE discretization with 96 edge sources
 331 per edge, that is, $96 \cdot 12 = 1152$ edge sources and $96 \cdot 12 \cdot 96 \cdot 6 = 663552$ unknowns in the
 332 \mathbf{q} -vector (eq. 9). ESIEBEM then used the same numbers of surface mesh elements and edge
 333 sources.

334 Fig. 3 shows the sound pressure level difference for the methods ESIEBEM and ESIE,
 335 relative to the reference results given by the BEM for 100 Hz and for 500 Hz. It can be

336 quite clearly seen that the ESIE has problems for some of the eight expected receiver angles,
 337 namely those that are very close to one of the infinite planes that contain the cube surfaces
 338 as mentioned in Section II B. Confirming one of the goals of developing the hybrid method,
 339 the ESIEBEM seems to give accurate results for those positions and therefore overcomes
 340 the singularities of the problematic receiver positions of the ESIE.

341 Comparing Fig. IV A and Fig. IV A, it can be observed that the sound pressure level
 342 difference between the two methods, ESIE and ESIEBEM, and the reference result given by
 343 BEM, is in much better agreement for 100 Hz than for 500 Hz. The mesh sizes are the same
 344 for both calculations so the number of elements per wavelength is five times higher in case
 345 (a) than in case (b).

346 Two receivers have been chosen and depicted in Fig. 3 for further study in the next
 347 sections: R_1 at 142.6° is a non-problematic receiver for ESIE and R_2 at 239.5° is a receiver
 348 close to a singularity of the ESIE propagation integral in Eq. 6.

349 B. Convergence for the three methods

350 Each method has been run for different discretizations and here, the convergence of each
 351 method towards their respective final value is analyzed further.

352 The potential for using extrapolation to find an estimate of the ultimate/final result, for
 353 an infinitely large number of elements, is explored below. This is the same technique as is
 354 used in Richardson extrapolation, where it is assumed that each computed value, p_n , based
 355 on a discretization step, Δh_n , is governed by a Taylor expansion around the final value,

$$p_n = p_{\text{final}} + C_0 \Delta h_n^{k_0} + \dots \quad (14)$$

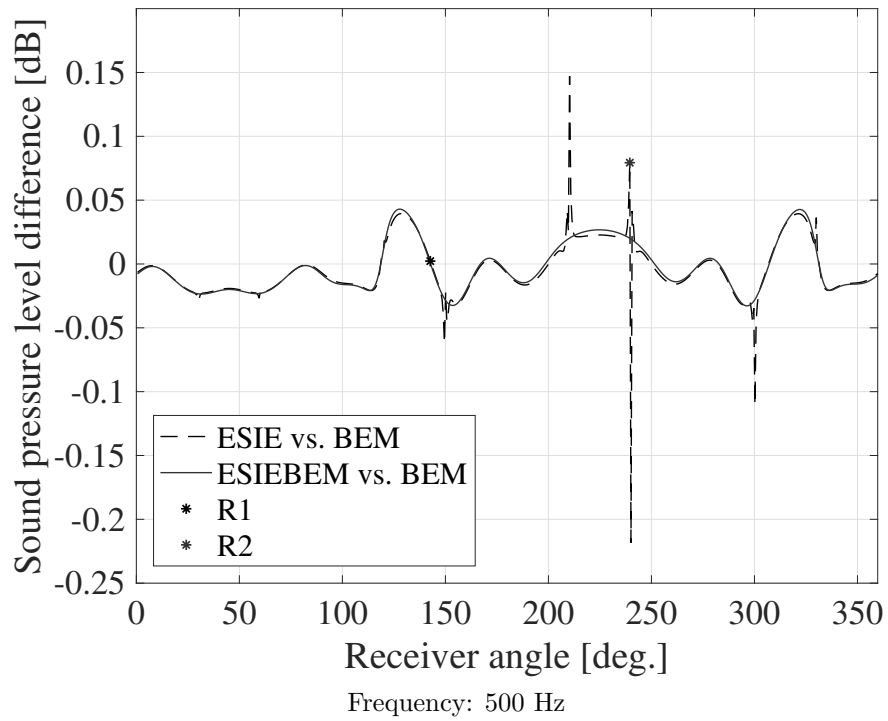
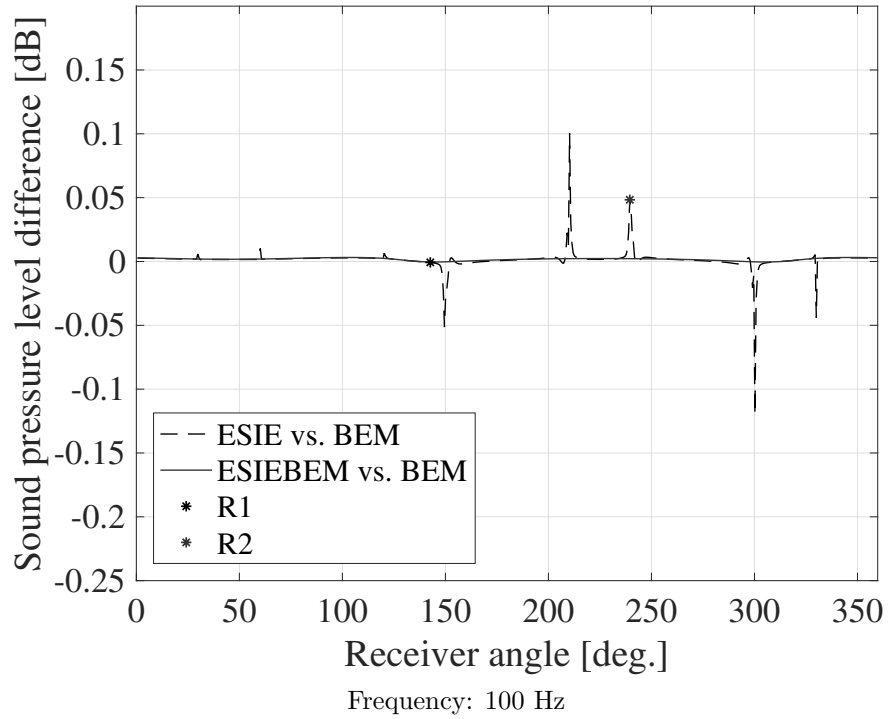


FIG. 3. Sound pressure level difference as function of the receiver angle for the 629 receivers in Fig. 2. Reference method is the BEM with 50872 elements. Two receiver positions are marked, R_1 and R_2 , that will be studied further. The frequency is: a) 100 Hz and b) 500 Hz.

356 where k_0 is a known or unknown exponent of the error convergence for the method at hand,
 357 and only the first polynomial term is kept. For the methods employed here, the sought value
 358 p_{final} can be found via a straight line fit to the data points p_n against $1/\text{BEM-mesh-size}$, for
 359 the BEM and ESIEBEM results and $1/\text{ESIE-edgesource-number}^2$ for the ESIE results. The
 360 final value would then be the intersection of the straight line with the y -axis representing a
 361 potential value for an infinite mesh of each method respectively.

362 Fig. 4 shows one example for the real part of the sound pressure amplitude at 100 Hz
 363 for all three methods, for receiver R_1 at 142.6° chosen in the previous section. The results
 364 are plotted versus $1/\text{BEM-mesh-size}$ for the BEM and ESIEBEM, and $1/\text{ESIE-edgesource-}$
 365 number^2 multiplied by 10^2 for visual purposes, respectively. The four finest discretizations
 366 for each method have been used for a straight line-fit, and the subsequent extrapolation to
 367 the y -axis crossing will be an estimate of the p_{final} values for each method.

368 It can be seen that the results for all three methods follow a trend that becomes rather
 369 linear for the finer discretizations, which supports the expected error convergence exponent:
 370 BEM and ESIEBEM converge as $\mathcal{O}(1/N)$ and ESIE as $\mathcal{O}(1/N^2)$. As a sidenote, it can
 371 be pointed out that a uniform discretization of the edges in the ESIE gives a $\mathcal{O}(1/N)$ -
 372 convergence, so the Gauss-Legendre quadrature used in the implementation gives a much
 373 better accuracy for practically the same computational cost.

374

375 An interesting way to evaluate the convergence of these results is presented here. Figs. 5
 376 and 6 show the trajectories of the complex sound pressure amplitude, as the discretization
 377 is refined, for all three methods, for the two different receivers R_1 and R_2 , for 100 Hz (Fig.

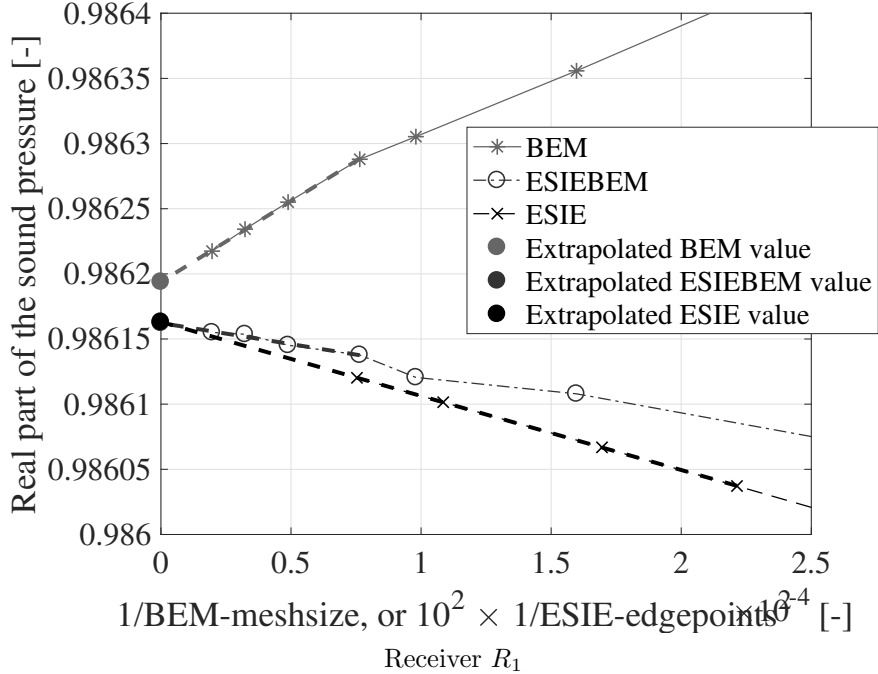


FIG. 4. Real part of the sound pressure amplitude at 100 Hz and receiver R_1 , computed with a number of different discretizations for the BEM, ESIEBEM, and ESIE method. The results for the four finest discretizations have been used for the linear regression.

378 5) and 500 Hz (Fig. 6), respectively. Also the extrapolated values obtained for each method
 379 have been plotted, except for the ESIE method at R_2 since this receiver is near a singularity
 380 location and its extrapolated value can not be determined.

381 It can be seen that for receiver R_1 the ESIE and ESIEBEM converge to its final value
 382 quite quickly and with a smooth and uniform trajectory, while the BEM takes larger steps.

383 For the ESIE problematic receiver, R_2 , it can be observed that the final value is not at
 384 the end of a smooth trajectory, since the results jump back and forth as the discretisation is
 385 increased. The singularity of the ESIE for those problematic positions might be solved by
 386 simply refining enough the discretization of the integration points. However, the convergence

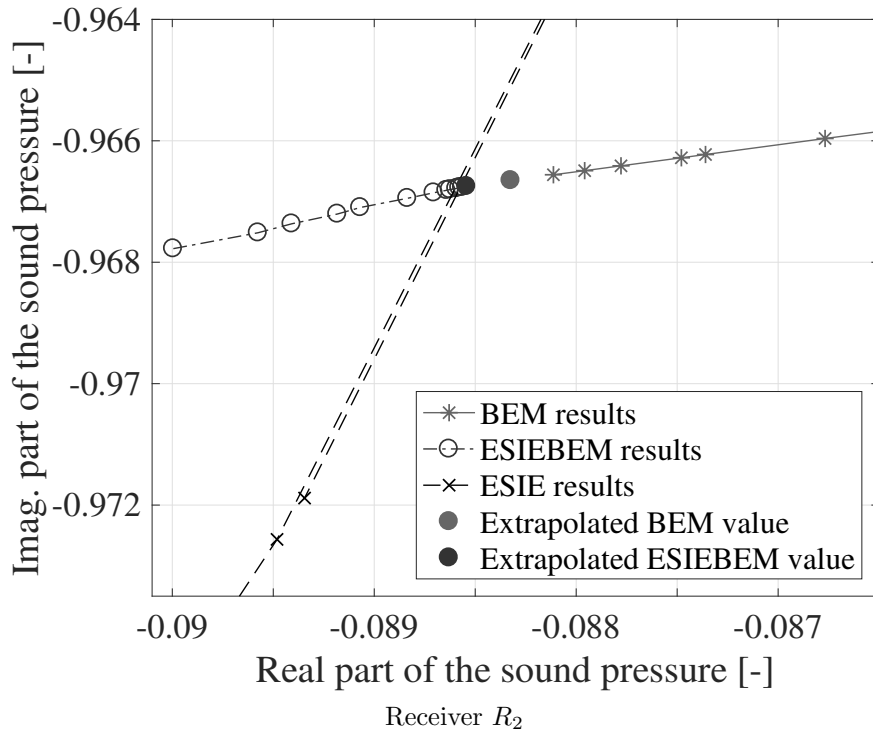
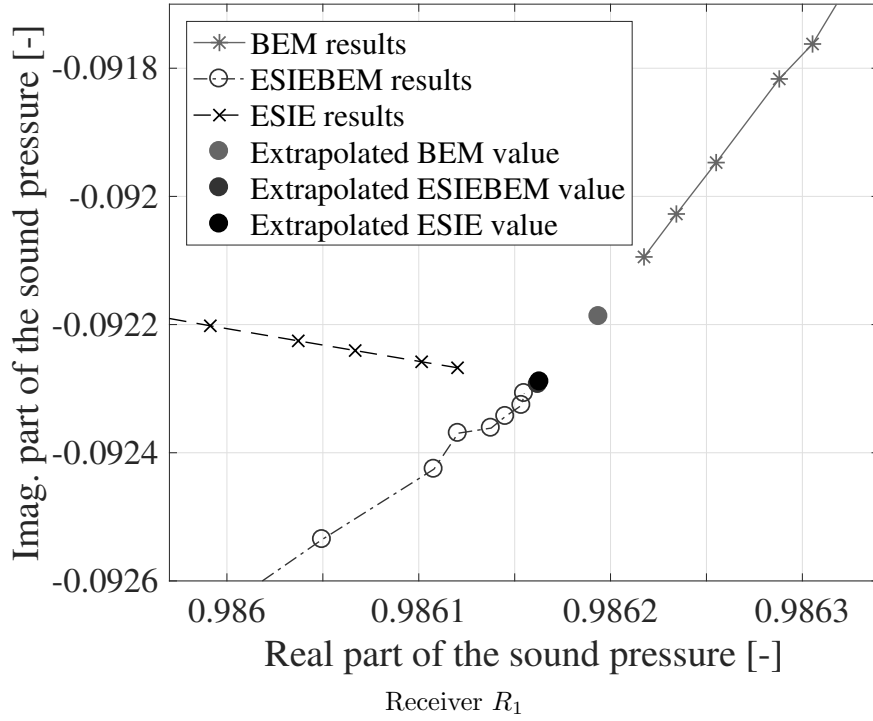


FIG. 5. Convergence for the BEM, ESIE and ESIEBEM for the cube test case at 100 Hz. The sound pressure amplitude is plotted in the complex plane and each point corresponds to a different mesh size or to a different number of edge sources, for a) receiver R_1 , and b) receiver R_2

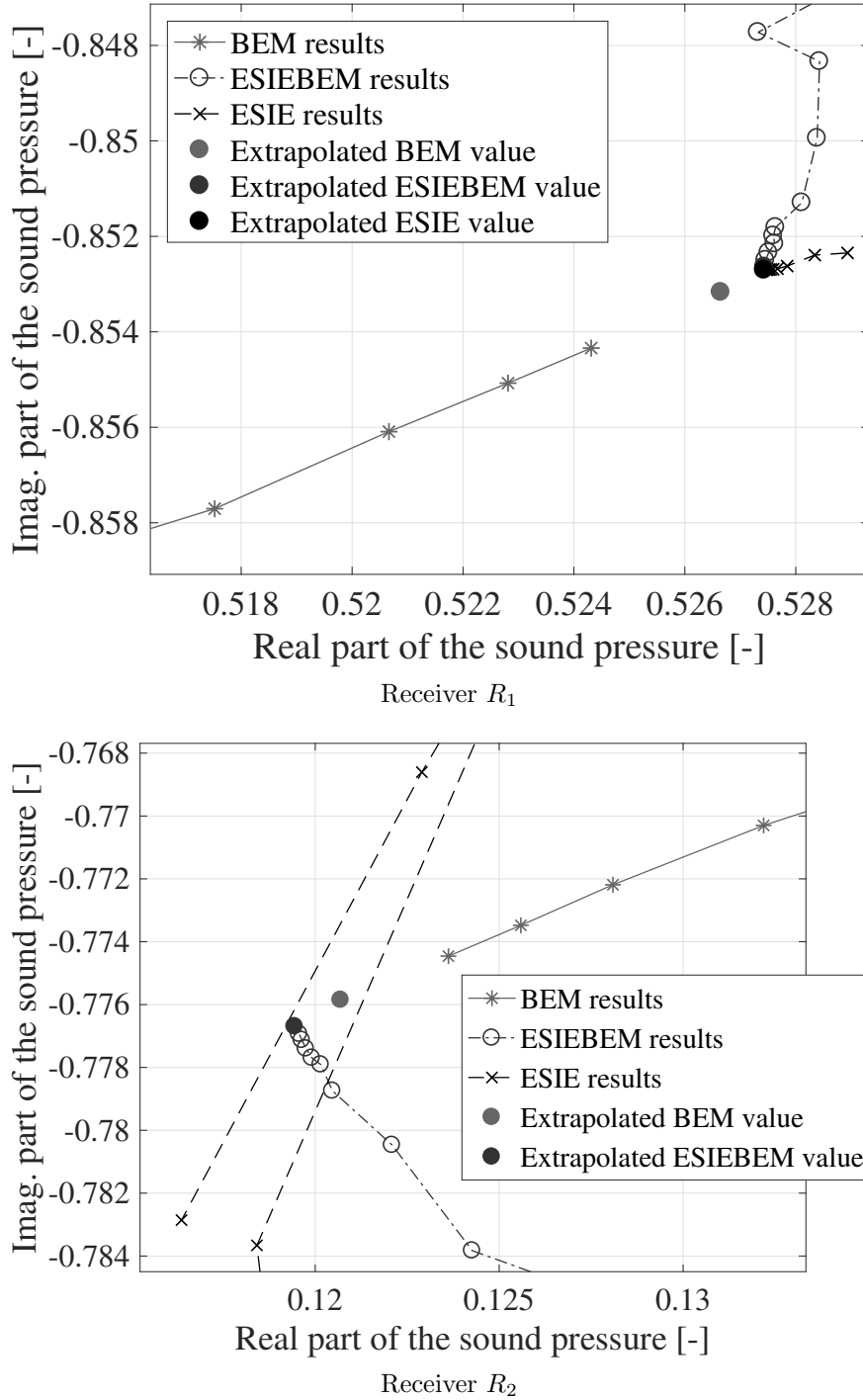


FIG. 6. Convergence for the BEM, ESIE and ESIEBEM for the cube test case at 500 Hz. The sound pressure is plotted in the complex plane and each point corresponds to a different mesh size or to a different number of edge sources, for a) receiver R_1 , and b) receiver R_2

387 is very slow and it is not clear that the refining would ultimately converge to the correct
388 result.

389 It is also interesting to notice that the extrapolated estimates of the result with the finest
390 meshes for each method are much closer to each other than the results for the finest dis-
391 cretizations themselves. Certainly, the value of p_{final} that is found in the way explained above
392 (see Eq. (14)) is not the ultimate result, but it can be viewed as a more accurate estimate
393 of the ultimate result than the best computed result, $p_{n,\text{max}}$ with the finest discretization.

394 Now, the convergence of each method towards their respective final value, p_{final} , can be
395 studied. For each receiver, the relative error is computed with this extrapolated estimate of
396 the final result used as reference result,

$$\epsilon_{\text{rel}}^i = \left| \frac{\hat{p}_i - p_{\text{ref}}}{p_{\text{ref}}} \right|, \quad (15)$$

397 where \hat{p}_i is the sound pressure at receiver i and p_{ref} the reference sound pressure, in this
398 case, $p_{\text{ref}} = p_{\text{final}}$. This relative error will vary among receiver positions, with potentially very
399 large variations, so here the median value, rather than the mean, across receiver positions
400 is presented as follows. Figs. IV B and IV B present the median relative error over all the
401 receivers as function of BEM-mesh elements and number of edge sources, for the frequencies
402 100 Hz and 500 Hz, respectively. The median has been chosen here rather than the mean
403 or the maximum absolute error to basically remove the well known inaccurate results given
404 by ESIE for certain receiver positions as discussed earlier.

405 All three methods display convergences with the assumed rates. Notably, the use of
406 the extrapolated result as a reference, rather than the result for the finest discretization,
407 makes these curves follow the trends very well. It is worth to mention that those rates would

408 increase in case polynomials of higher order in BEM or ESIEBEM were used. Apparently, the
 409 ESIE converges to the extrapolated value with the fastest rate and the ESIEBEM converges
 410 to its extrapolated value with smaller error than the BEM. Thus, the hybrid ESIEBEM
 411 method developed in this paper, is performing very well compared to the boundary element
 412 method. It is interesting to note that the ESIE results reach similar accuracies for both
 413 frequencies, 100 Hz and 500 Hz. The error of BEM and ESIEBEM, on the other hand, gets
 414 one order of magnitude higher from the results at 100 Hz to the ones at 500 Hz, which is
 415 not surprising since the discretizations are the same for both frequencies.

416 It should also be realized that the reference result in Fig. 7 was computed from the
 417 results of each method to demonstrate the method's convergence. The studied acoustic
 418 scattering problem does not have an analytical solution, but the extrapolation of the BEM
 419 can be considered as a reference result to compare the results of the methods. Fig. 8 shows
 420 different measures of the relative error (over all receivers) using the BEM extrapolation value
 421 as a reference p_{ref} in Eq. 15, in this case, $p_{\text{ref}} = p_{\text{final,BEM}}$, for 100 Hz (Fig. IV B) and 500 Hz
 422 (Fig. IV B): the maximum error, the 90% percentile, the mean error and the 50% percentile
 423 (the median) over all 629 receivers for each of the three methods. First, it is interesting to
 424 notice the significant difference between the maximum error and the 90% percentile for the
 425 ESIE which is due to the well known problematic receivers. This effect causes the mean
 426 and the median errors for the ESIE to be quite different too, and as mentioned earlier, that
 427 is the reason to make use of the median rather than the mean in Fig. 7. For ESIEBEM
 428 and BEM, both the maximum error and the 90% percentile are quite close, as well as the
 429 median and the mean errors.

430 It can also be observed that the ESIE and ESIEBEM have smaller errors relative to the
431 BEM extrapolation than the proper BEM up to a certain fine BEM mesh. The relative
432 errors of the ESIE and ESIEBEM typically (90% quartile) reach $4 \cdot 10^{-4}$ (ESIE) and $3 \cdot 10^{-4}$
433 (ESIEBEM) at 100 Hz, and $2 \cdot 10^{-3}$ at 500 Hz, relative to the extrapolated BEM result,
434 which is considered as the best possible reference result.

435 V. CONCLUSIONS

436 A new hybrid method (ESIEBEM) combining a technique which is an extension of
437 geometrical-acoustics, the edge source integral equation method (ESIE), and a wave-based
438 technique, the boundary element method (BEM), has been introduced in this paper. A
439 benchmark case has been presented to study the performance of the proposed ESIEBEM
440 method: the study of the scattering by a rigid cube for plane wave incidence; with receivers
441 around the cube, not very close to the cube surface.

442 The results obtained by the ESIEBEM have been compared with those given by the
443 boundary element method (BEM) and the edge diffraction based method (ESIE). Compu-
444 tations have been carried out with several different discretizations, and linear extrapolation
445 has been employed to estimate more accurate results than the computed ones. The ES-
446 IE BEM inherits the property of the ESIE to give accurate results for convex bodies but has
447 the advantage that it overcomes the singularities of the ESIE for certain receiver positions.
448 Resonances in the reciprocal interior problem are not related to the ESIEBEM and CHIEF
449 points are not needed.

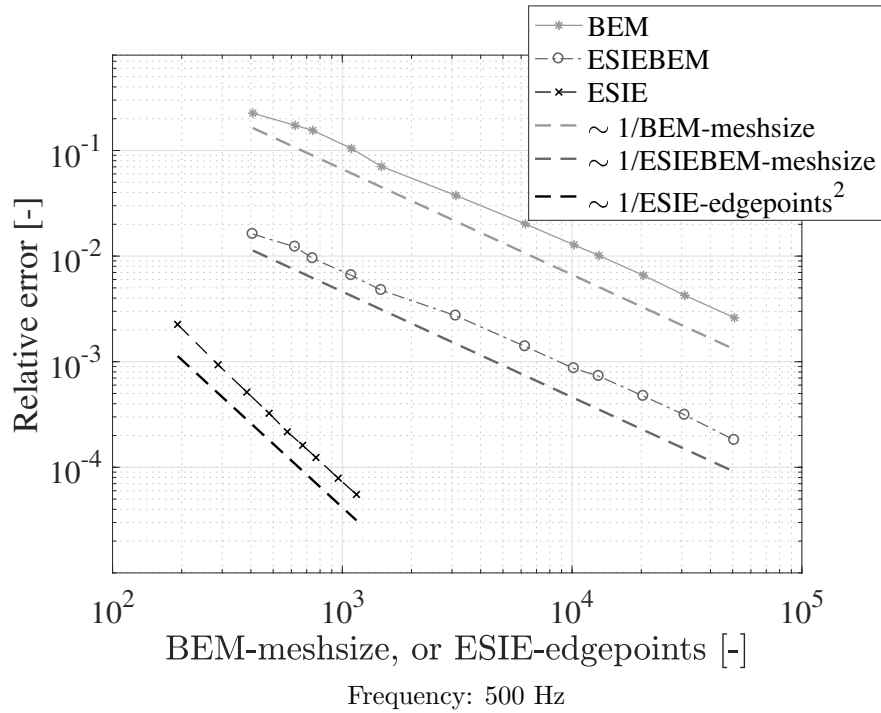
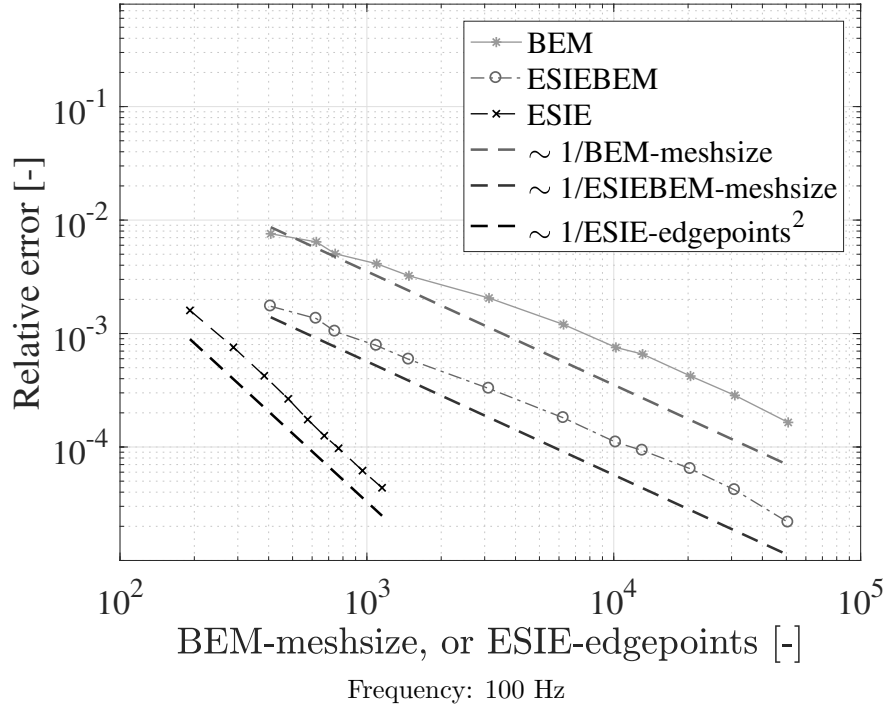


FIG. 7. Median relative error, across all 629 receivers, as a function of number of nodes in the mesh (for the BEM and ESIEBEM) or number of edge integration points (for the ESIE). The reference result is the linear regression extrapolation for each method respectively. Calculated frequencies: a) 100 Hz and b) 500 Hz.

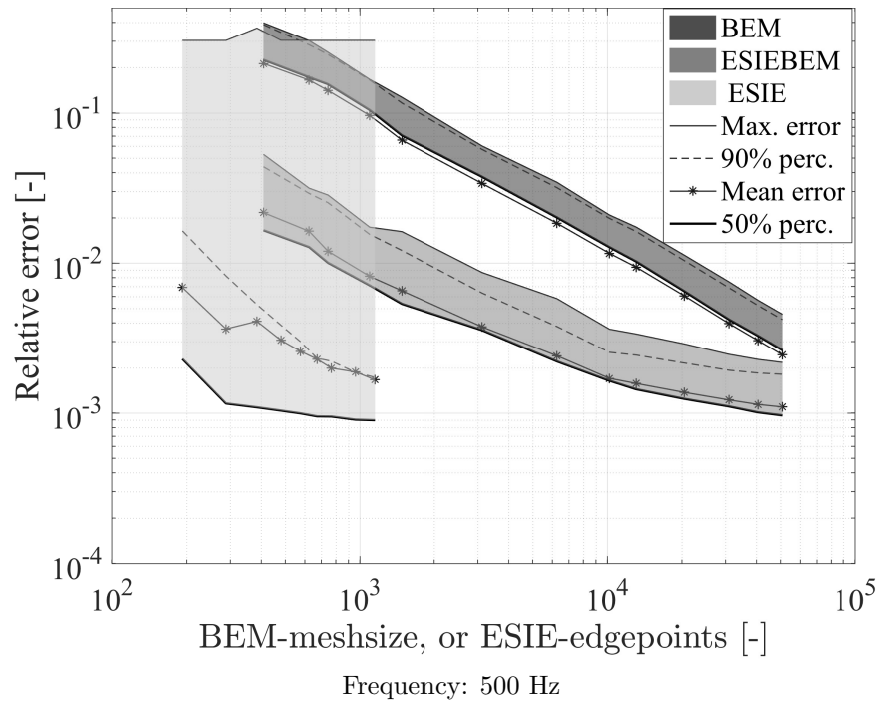
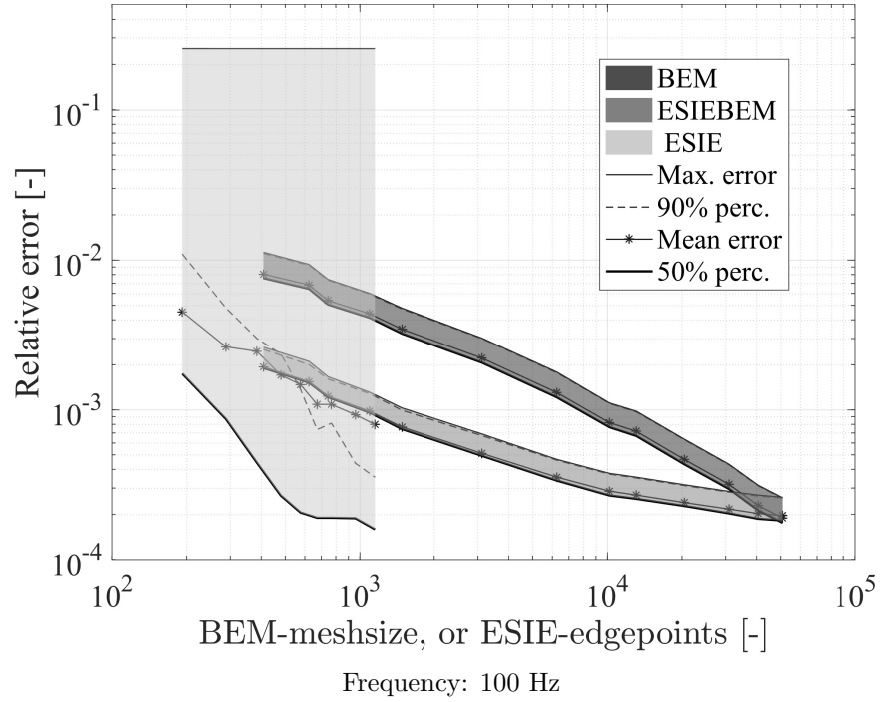


FIG. 8. Relative error, across all 629 receivers, as a function of number of nodes in the mesh (for the BEM and ESIEBEM) or number of edge integration points (for the ESIE). The reference result is the linear regression extrapolation of BEM for all three methods. The frequency is a) 100 Hz and b) 500 Hz.

450 The computational time for the ESIE and ESIEBEM seems advantageous compared to
451 the BEM for the cube test case. The ESIEBEM also shows good accuracy compared to the
452 BEM and the ESIE. The convergence for two receivers was investigated in detail and the
453 ESIEBEM converges to the similar value as the ESIE. The relative errors of the ESIE and
454 ESIEBEM typically (90% quartile) reach $4 \cdot 10^{-4}$ (ESIE) and $3 \cdot 10^{-4}$ (ESIEBEM) at 100 Hz,
455 and $2 \cdot 10^{-3}$ at 500 Hz, relative to the extrapolated BEM result, which is considered as the
456 best possible reference result.

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468 **VII. APPENDIX A: DIRECTIVITY FUNCTION FOR RECEIVERS ON THE**
469 **SURFACE**

470 The directivity function β in Eq. (3), also used in Eq.(4), is defined by

$$\beta(R, z, S) = \sum_{i=1}^4 \frac{\sin(\nu\theta_i)}{\cosh(\nu\eta) - \cos(\nu\theta_i)}, \quad (16)$$

471 where the angles θ_i are

$$\theta_1 = \pi + \theta_S + \theta_R, \quad \theta_2 = \pi - \theta_S + \theta_R, \quad (17)$$

$$\theta_3 = \pi + \theta_S - \theta_R, \quad \theta_4 = \pi - \theta_S - \theta_R, \quad (18)$$

472 and η is an auxiliary function defined by

$$\eta = \cosh^{-1} \left(\frac{\cos \varphi_S \cos \varphi_R + 1}{\sin \varphi_S \sin \varphi_R} \right). \quad (19)$$

473 When the receiver R is placed on the surface, so $\theta_R = 0$, then $\theta_1 = \theta_3$ and $\theta_2 = \theta_4$. Therefore,

474 Eq. (16) is simplified to

$$\beta(R, z, S) = 2 \cdot \left(\frac{\sin(\nu\pi + \nu\theta_S)}{\cosh(\nu\eta) - \cos(\nu\pi + \nu\theta_S)} + \frac{\sin(\nu\pi - \nu\theta_S)}{\cosh(\nu\eta) - \cos(\nu\pi - \nu\theta_S)} \right), \quad (20)$$

475 By using the trigonometric identities

$$\sin(\nu\pi - \nu\theta_S) = \sin(\nu\pi) \cos(\nu\theta_S) - \cos(\nu\pi) \sin(\nu\theta_S), \quad \sin(\nu\pi + \nu\theta_S) = \sin(\nu\pi) \cos(\nu\theta_S) + \cos(\nu\pi) \sin(\nu\theta_S), \quad (21)$$

$$\cos(\nu\pi - \nu\theta_S) = \cos(\nu\pi) \cos(\nu\theta_S) + \sin(\nu\pi) \sin(\nu\theta_S), \quad \cos(\nu\pi + \nu\theta_S) = \cos(\nu\pi) \cos(\nu\theta_S) - \sin(\nu\pi) \sin(\nu\theta_S), \quad (22)$$

476 and the equality

$$\sin(\nu\pi) \cos(\nu\pi) = \frac{\sin(2\nu\pi)}{2}, \quad (23)$$

477 Eq. (20) can be rewritten as

$$\beta(R, z, S) = 4 \cdot \left(\frac{\sin(\nu\pi) \cos(\nu\theta_S) \cosh(\nu\eta) - \frac{\sin(2\nu\pi)}{2}}{(\cosh^2(\nu\eta) - 2 \cosh(\nu\eta) \cos(\nu\pi) \cos(\nu\theta_S) + \cos(\nu(\pi - \theta_S)) \cos(\nu(\pi + \theta_S)))} \right). \quad (24)$$

478 This expression is used in the ESIEBEM to compute the first order diffraction component
 479 in Eq. (3) at the collocation points on the scatterer's surface. Interestingly enough, the
 480 directivity function $\beta(R, z_1, z_2)$ used to obtain the higher order diffraction term in Eq. (6)
 481 can also be reduced from Eq. (24). Since z_2 is on the surface, $\theta_S = \theta_{z_2} = 0$ and the simplified
 482 expression for $\beta(R, z_1, z_2)$ turns to be

$$\beta(R, z_1, z_2) = 4 \cdot \left(\frac{\sin(\nu\pi) \cosh(\nu\eta) - \frac{\sin(2\nu\pi)}{2}}{(\cosh(\nu\eta) - \cos(\nu\pi))^2} \right), \quad (25)$$

483 where η is defined in Eq. (19).

484 **VIII. APPENDIX B: THE SPARSENESS OF THE H-MATRIX FOR THE ESIE**

485 The sparsity of the \mathbf{H} -matrix in Eq. (11) can be understood as follows. For simplicity
 486 a cube is chosen as scattering object such that all the edges have the same length. The
 487 number of discretization points per edge is denoted with $N_{\text{per edge}}$. Each edge point can
 488 then see $6N_{\text{per edge}}$ other edge points, because each edge can see exactly 6 of the 12 edges of
 489 the cube. Other polyhedral shapes than the cube will have other values than 6/12. There
 490 are altogether 12 edges and consequently a total of $N_{\text{total}} = 12N_{\text{per edge}}$ edge discretization
 491 points. The number of unknowns (edge source amplitudes $q(z_1, z_2)$) is then $N_{\text{unknowns}} =$
 492 $12N_{\text{per edge}} \cdot 6N_{\text{per edge}} = 0.5N_{\text{total}}^2$. Thus, the vector \mathbf{q} and the transfer matrix \mathbf{H} will
 493 have sizes of $[0.5N_{\text{total}}^2, 1]$ and $[0.5N_{\text{total}}^2, 0.5N_{\text{total}}^2]$ respectively. Each row in this \mathbf{H} -matrix
 494 will obviously have $0.5N_{\text{total}}^2$ elements, but since each edge source can be reached only by
 495 $0.5N_{\text{total}}$ other edge sources, only $0.5N_{\text{total}}$ entries in each row will be non-zero. Altogether,
 496 the \mathbf{H} -matrix has $0.25N_{\text{total}}^3$ non-zero elements of all its $0.25N_{\text{total}}^4$, which represents a high
 497 degree of sparseness.

498 The computational cost for obtaining the edge source amplitudes $\mathbf{q}_{\text{final}}$ is given by setting
 499 up the \mathbf{H} -matrix and the iterative solution of Eq. (11), so the calculation time, T , will be

$$\begin{aligned}
 T_{\mathbf{q}_{\text{final}}} &= C_{\text{set-up}}N_{\text{total}}^3 + C_{\text{iter.}}N_{\text{truncation}}N_{\text{total}}^3 \\
 &\sim (1 + C_{\text{iter.,rel.}}N_{\text{truncation}})N_{\text{total}}^3 \quad (26)
 \end{aligned}$$

500 where the various C are constants, and the value of $N_{\text{truncation}}$ is typically below 20^{23} .

501 Finally, the higher-order diffraction sound pressure at $N_{\text{receivers}}$ receiver points is obtained
502 by calculating the double integral in Eq. (6), which is computed for the same discretization
503 as described above. The time for this stage will be

$$T_{\text{prop.}} \sim N_{\text{receivers}} N_{\text{unknowns}} \sim N_{\text{receivers}} N_{\text{total}}^2. \quad (27)$$

504 As indicated by Eqs. (26) and (27), the calculation of the $\mathbf{q}_{\text{final}}$ is typically the dominating
505 computational stage, but for the application here, the number of receiver points is high, so
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