Fault Tolerant Control of a Proton Exchange Membrane Fuel Cell Using Takagi-Sugeno Virtual Actuators

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Abstract

In this paper, a fault tolerant control (FTC) strategy for proton exchange membrane (PEM) fuel cells based on the use of virtual actuators and the Takagi-Sugeno (TS) approach is proposed. The overall solution relies on adding a virtual actuator in the control loop to hide the fault from the controller point of view, allowing it to see the same plant as before the fault, such that the stability and some desired performances are preserved. The proposed methodology is based on the use of a reference model, where the resulting nonlinear error model is brought to a Takagi-Sugeno form using a gridding approach. The TS model is suitable for designing a controller using linear matrix inequalities (LMI)-based techniques, such that the resulting closed-loop error system is stable with poles placed in some desired region of the complex plane. Simulation results are used to show the effectiveness of the proposed approach.

Keywords: Takagi-Sugeno model, Virtual actuator, Reference model based control, Gain-scheduling, PEM Fuel Cell, LMIs.

1. Introduction

Proton exchange membrane (PEM, also known as polymer electrolyte membrane) fuel cells are electrochemical devices which directly convert the chemical energy of hydrogen into electrical energy [?]. Fuel cell systems offer a clean alternative to energy production and are a very active research field because of many possible applications in distributed generation solutions [?]. A good performance of these devices is closely related to the kind of control that is used, so a study of different control alternatives is considered in [?]. A fuel cell integrates many components into a power system, which supplies electricity to an electric load or to the grid. Several devices, such as DC/DC or DC/AC

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converters, batteries or ultracapacitors, are included in the system and, in case the fuel cell is not fed directly with hydrogen, a reformer must also be used. Therefore, there are many control loops schemes depending on the devices that must be controlled. The lower control level takes care of the main control loops inside the fuel cell, which are basically fuel/air feeding, humidity, pressure and temperature. The upper control level is in charge of the whole system, integrating the electrical conditioning, storage and reformer (if necessary). Many control strategies have been proposed in the recent literature, e.g. optimal control [?], model predictive control [?] and sliding mode control [?], and several efforts have been put in providing an accurate model for this kind of systems [??].

Since fuel cells are very complex systems, they are vulnerable to faults that can cause their stop or their permanent damage [??]. Hence, it is interesting to add some fault tolerant capabilities to the control system, in order to maintain the fuel cell operating even in the presence of faults [??]. Fault tolerant control (FTC) systems are able to maintain desirable closed-loop performance, or with an acceptable degradation, and preserve stability conditions in the presence of component and/or instrument faults [??]. The existing FTC design techniques can be classified into passive and active approaches (see [?] for a review). Passive FTC techniques exploit the fact that within certain margins, the control law has inherent fault tolerance capabilities, allowing the system to cope with the fault presence, while the active FTC techniques compensate the faults either by selecting a precalculated control law or by synthesizing online a new control strategy. The adaptation of the control law is done using some information about the fault so as to satisfy the control objectives with minimum performance degradation after the fault occurrence.

In recent years, the fault-hiding paradigm has been proposed as an active strategy to obtain fault tolerance [?]. In this paradigm, the controller reconfiguration (CR) unit reconfigures the faulty plant instead of the controller/observer. The nominal controller is kept in the loop by inserting a reconfiguration block between the faulty plant and the nominal controller/observer when a fault occurs. The reconfiguration block is chosen so as to hide the fault from the controller point of view, allowing it to see the same plant as before the fault. The reconfiguration block is named virtual actuator in case of actuator faults and virtual sensor in case of sensor faults. The virtual actuator strategy has been initially proposed in a state space form for LTI systems [?], and successfully extended to linear parameter varying (LPV) [?], Takagi-Sugeno (TS) [?] and piecewise affine [?] systems. An equivalent formulation in input-output form has been proposed in [?].

Recently, the complex and nonlinear dynamics of the power generation systems based on fuel cell technology, described in detail in [?], led to the use of linear models that include parameters varying with the operating point (known as LPV models) not only for advanced control techniques [?] but also for model-based fault diagnosis algorithms [?]. As an alternative to the LPV paradigm, TS systems, as introduced by [?], provide an effective way of representing nonlinear systems with the aid of fuzzy sets, fuzzy rules and a set of

local linear models. The overall model is obtained by merging the local models through fuzzy membership functions. In some recent works, the TS paradigm has been successfully applied to the PEM fuel cell system for solving the problems of modeling [?], fault diagnosis [?] and state observation [?].

In this paper, an FTC strategy based on the use of virtual actuators and a TS modeling approach for PEM fuel cells is proposed. The fault tolerant methodology is based on the use of a reference model, where the resulting nonlinear error model is brought to a TS form using a mix of sector nonlinearity and gridding approaches. The TS model is suitable for designing the controller and the virtual actuators, which hide the faults allowing to achieve fault tolerance, using linear matrix inequalities (LMI)-based techniques, such that the resulting closed-loop error system is stable with poles placed in some desired region of the complex plane. Simulation results are used to show the effectiveness of the proposed approach.

The structure of the paper is the following: Section 2 describes the PEM Fuel Cell and presents its nonlinear model. Section 3 shows how, using a reference model, a TS error model, suitable for designing a TS controller using LMI-based techniques, can be obtained. Section 4 presents the proposed FTC strategy based on virtual actuators. The application of the presented theory to a PEM Fuel Cell case study is given in Section 5 and the results obtained in simulation are presented in Section 6. Finally, the main conclusions are outlined in Section 7.

2. Description and modeling of PEM fuel cells

2.1. PEM fuel cell description

A fuel cell is an electrochemical energy converter that transforms the chemical energy of fuel into electrical current. It has an electrolyte, a negative electrode and a positive electrode, and it generates direct electrical current through an electrochemical reaction. Typical reactants for fuel cells are hydrogen as fuel and oxygen as oxidant that, once the reaction takes place, produce water and waste heat.

The basic physical structure of a fuel cell consists of an electrolyte layer in contact with a porous anode and cathode electrode plates. There are different kinds of electrolyte layers. Here, a PEM fuel cell is used as a case study. The PEM has a special property: it conducts protons but is impermeable to gas (the electrons are blocked through the membrane). Auxiliary devices are required to ensure the proper operation of the fuel cell stack: an air compressor, a hydrogen tank, a supply manifold and a return manifold.

2.2. PEM fuel cell model

The model used in this work has been presented in [?] and is widely accepted in the control community as a good representation of the behavior of a fuel cell stack (FCS) system. Throughout this work, it is assumed that local loops control the compressor speed ω_{cp} in order to achieve the desired

compressor mass flow W_{cp} , and the hydrogen supplied to the anode m_{H_2} . Hence, the proposed fault tolerant methodology will be applied to the subsystem made up by the supply manifold, the return manifold and the cathode.

The supply manifold includes pipe and stack manifold volumes between the compressor and the fuel cells, and is governed by mass continuity and energy conservation equations [?]:

$$\dot{m}_{sm} = W_{cp} - k_{sm,out} \left[p_{sm} - \frac{(m_{O_2} R_{O_2} + m_{N_2} R_{N_2}) T_{st}}{V_{ca}} \right]$$
(1)

$$\dot{p}_{sm} = \frac{\gamma R_a}{V_{sm}} \left\{ W_{cp} \left[T_{atm} + \frac{T_{atm}}{\eta_{cp}} \left(\left(\frac{p_{sm}}{p_{atm}} \right)^{\frac{\gamma - 1}{\gamma}} - 1 \right) \right] - k_{sm,out} T_{sm} \left[p_{sm} - \frac{(m_{O_2} R_{O_2} + m_{N_2} R_{N_2}) T_{st}}{V_{ca}} \right] \right\}$$
(2)

where:

$$T_{sm} = \frac{p_{sm}V_{sm}}{m_{sm}R_a} \tag{3}$$

The return manifold pressure is governed by mass conservation and the ideal gas law through isothermal assumptions [?]:

$$\dot{p}_{rm} = \frac{R_a T_{rm}}{V_{rm}} k_{ca,out} \left(\frac{(m_{O_2} R_{O_2} + m_{N_2} R_{N_2}) T_{st}}{V_{ca}} - p_{rm} \right) - \frac{R_a T_{rm}}{V_{rm}} k_{rm,out} \left(p_{rm} - p_{atm} \right)$$
(4)

The cathode flow dynamics is described by the following differential equations:

$$\dot{m}_{O_2} = \chi_{O_2,i} k_{sm,out} \left(p_{sm} - \frac{(m_{O_2} R_{O_2} + m_{N_2} R_{N_2}) T_{st}}{V_{ca}} \right) - \chi_{O_2,o} k_{ca,out} \left(\frac{(m_{O_2} R_{O_2} + m_{N_2} R_{N_2}) T_{st}}{V_{ca}} - p_{rm} \right) - \frac{M_{O_2} n I_{st}}{4F}$$
(5)

$$\dot{m}_{N_2} = (1 - \chi_{O_2,i}) k_{sm,out} \left(p_{sm} - \frac{(m_{O_2} R_{O_2} + m_{N_2} R_{N_2}) T_{st}}{V_{ca}} \right) - (1 - \chi_{O_2,o}) k_{ca,out} \left(\frac{(m_{O_2} R_{O_2} + m_{N_2} R_{N_2}) T_{st}}{V_{ca}} - p_{rm} \right)$$
(6)

with:

$$\chi_{O_2,i} = \frac{y_{O_2,i} M_{O_2}}{y_{O_2,i} M_{O_2} + (1 - y_{O_2,i}) M_{N_2}}$$
(7)

$$\chi_{O_2,o} = \frac{y_{O_2,o} M_{O_2}}{y_{O_2,o} M_{O_2} + (1 - y_{O_2,o}) M_{N_2}} \tag{8}$$

$$y_{O_2,o} = \frac{m_{O_2} R_{O_2}}{m_{O_2} R_{O_2} + m_{N_2} R_{N_2}}$$
(9)

The model used for control purposes presented in the following has three

state variables, i.e. p_{sm} , p_{rm} and w_{ca} , defined as:

$$w_{ca} = m_{O_2} R_{O_2} + m_{N_2} R_{N_2} \tag{10}$$

two control inputs, i.e. W_{cp} and $k_{rm,out}$, and an exogenous input, I_{st} , that corresponds to the load, and acts as a disturbance, that can be included in the reference model in order to generate an appropriate feedforward action.

Notice that the following state equation can be obtained from (10), taking into account (5) and (6):

$$\dot{w}_{ca} = k_{sm,out} \left(p_{sm} - w_{ca} \frac{T_{st}}{V_{ca}} \right) \left[\chi_{O_2,i} R_{O_2} + (1 - \chi_{O_2,i}) R_{N_2} \right] - k_{ca,out} \left(w_{ca} \frac{T_{st}}{V_{ca}} - p_{rm} \right) \left[\chi_{O_2,o} R_{O_2} + (1 - \chi_{O_2,o}) R_{N_2} \right]$$
(11)

Four sensors are available, measuring p_{sm} , p_{rm} , T_{sm} and the pressure in the cathode p_{ca} , related to w_{ca} by:

$$p_{ca} = w_{ca} \frac{T_{st}}{V_{ca}} \tag{12}$$

The efficiency optimization of the current system can be achieved by regulating the oxygen mass inflow towards the stack cathode [?]. If an adequate oxidant flow is ensured through the stack, the load demand is satisfied with minimum fuel consumption. In addition, oxygen starvation and irreversible damage are averted. To accomplish such an oxidant flow is equivalent to maintaining at a suitable value the oxygen stoichiometry, defined as:

$$\lambda_{O_2} = \frac{k_{sm,out} \left[p_{sm} - \left(m_{O_2} R_{O_2} + m_{N_2} R_{N_2} \right) \frac{T_{st}}{V_{ca}} \right]}{\frac{M_{O_2} n I_{st}}{4F}}$$
(13)

3. Model reference control using Takagi-Sugeno techniques

3.1. Takagi-Suqeno modeling

Let us consider that the nonlinear model of the PEM fuel cell can be described by a TS model, which uses a set of local models merged together using fuzzy IF-THEN rules [?], as follows:

$$IF \ \vartheta_1(k) \ is \ M_{i1} \ AND \dots AND \ \vartheta_p(k) \ is \ M_{ip}$$

$$THEN \ x_i(k+1) = A_i x(k) + B_i u(k) \quad i = 1, \dots, N$$

$$(14)$$

where M_{ij} denote the fuzzy sets and N is the number of model rules; $x(k) \in \mathbb{R}^{n_x}$ is the state vector of the overall system, $x_i(k) \in \mathbb{R}^{n_x}$ is the state vector of the *i*th local model, $u(k) \in \mathbb{R}^{n_u}$ is the input vector, while A_i and B_i are matrices of appropriate dimensions. $\vartheta_1(k), \ldots, \vartheta_p(k)$ are premise variables that can be functions of exogenous variables, endogenous variables (states and/or inputs)

and/or time. Each linear consequent equation represented by $A_i x(k) + B_i u(k)$ is called a *subsystem*.

Given a pair (x(k), u(k)), the state of the TS system can easily be inferred by:

$$x(k+1) = \sum_{i=1}^{N} \rho_i(\vartheta(k)) (A_i x(k) + B_i u(k))$$
 (15)

where $\vartheta(k) = [\vartheta_1(k), \dots, \vartheta_p(k)]^T$ is the vector containing the premise variables, and $\rho_i(\vartheta(k))$ is defined as follows:

$$\rho_i(\vartheta(k)) = \frac{w_i(\vartheta(k))}{\sum_{i=1}^{N} w_i(\vartheta(k))}$$
(16)

$$w_i(\vartheta(k)) = \prod_{i=1}^{p} M_{ij}(\vartheta_j(k))$$
(17)

where $M_{ij}(\vartheta_j(k))$ is the grade of membership of $\vartheta_j(k)$ in M_{ij} and $\rho_i(\vartheta(k))$ is such that:

$$\begin{cases}
\sum_{i=1}^{N} \rho_i(\vartheta(k)) = 1 \\
\rho_i(\vartheta(k)) \ge 0, \quad i = 1, \dots, N
\end{cases}$$
(18)

3.2. Takaqi-Suqeno reference model and control law

For the synthesis of the TS controller, the following TS reference model is considered:

IF
$$\vartheta_1(k)$$
 is M_{i1} AND ... AND $\vartheta_p(k)$ is M_{ip}
THEN $x_{ref,i}(k+1) = A_i x_{ref}(k) + B_i u_{ref}(k)$
 $i = 1, ..., N$ (19)

where $x_{ref}(k) \in \mathbb{R}^{n_x}$ is the reference state vector of the overall system, $x_{ref,i} \in \mathbb{R}^{n_x}$ is the reference state vector of the *i*th local model, and $u_{ref}(k) \in \mathbb{R}^{n_u}$ is the reference input vector (feedforward action). The reference model generates the trajectory to be tracked by the real system. Given a pair $(x_{ref}(k), u_{ref}(k))$, the reference state can be inferred as:

$$x_{ref}(k+1) = \sum_{i=1}^{N} \rho_i (\vartheta(k)) (A_i x_{ref}(k) + B_i u_{ref}(k))$$
 (20)

Thus, considering the error, defined as $e(k) \triangleq x_{ref}(k) - x(k)$ and the new input $\Delta u(k) \triangleq u_{ref}(k) - u(k)$ (feedback action), the following TS error system is obtained:

$$e(k+1) = \sum_{i=1}^{N} \rho_i \left(\vartheta(k) \right) \left[A_i e(k) + B_i \Delta u(k) \right]$$
 (21)

The error system (21) is controlled through a TS error-feedback controller, defined as follows:

$$IF \ \vartheta_1(k) \ is \ M_{i1} \ AND \ \dots \ AND \ \vartheta_p(k) \ is \ M_{ip}$$

$$THEN \ \Delta u_{c,i}(k) = K_i e(k) \quad i = 1, \dots, N$$

$$(22)$$

such that the control action to be applied is inferred as the weighted mean:

$$\Delta u(k) = \sum_{i=1}^{N} \rho_i \left(\vartheta(k) \right) \Delta u_{c,i}(k)$$
 (23)

3.3. Controller design using an LMI-based approach

The TS controller (23) is designed using an LMI-based approach, where the desired specifications (in this case, stability and pole clustering) are guaranteed using the results from the quadratic Lyapunov framework [?]. Despite the introduction of conservativeness with respect to other existing approaches (e.g. those where the Lyapunov function is allowed to be fuzzy), the quadratic approach has undeniable advantages in terms of computational complexity.

In particular, the TS error system (21) with the error-feedback control law (23) is quadratically stable if and only if there exist $X = X^T > 0$ and K_i such that:

$$\begin{pmatrix} -X & (A_i + B_j K_i) X \\ X (A_i + B_j K_i)^T & -X \end{pmatrix} < 0$$
 (24)

for i, j = 1, ..., N.

On the other hand, pole clustering is based on the results obtained by [?], where subsets \mathcal{D} of the complex plane, referred to as *LMI regions*, are defined as:

$$\mathcal{D} = \{ z \in \mathbb{C} : f_{\mathcal{D}}(z) < 0 \} \tag{25}$$

where $f_{\mathcal{D}}$ is the *characteristic function*, defined as:

$$f_{\mathcal{D}}(z) = \alpha + z\beta + \bar{z}\beta^T = [\alpha_{hl} + \beta_{hl}z + \beta_{lh}\bar{z}]_{h,l \in [1,m]}$$
(26)

where $\alpha = \alpha^T \in \mathbb{R}^{m \times m}$ and $\beta \in \mathbb{R}^{m \times m}$. Hence, the TS error system (21) with error-feedback control law (23) has poles in \mathcal{D} if there exist $X_{\mathcal{D}} = X_{\mathcal{D}}^T > 0$ and K_i such that:

$$\left[\alpha_{hl}X_{\mathcal{D}} + \beta_{hl}\left(A_i + B_jK_i\right)X_{\mathcal{D}} + \beta_{lh}X_{\mathcal{D}}\left(A_i + B_jK_i\right)^T\right] \underset{h,l \in [1,m]}{<} 0 \tag{27}$$

for i, j = 1, ..., N.

Conditions (24)-(27) are bilinear matrix inequalities (BMIs) that can be brought to LMI form by using a single Lyapunov matrix $X = X_{\mathcal{D}} > 0$ and through the change of variables $\Gamma_i \triangleq K_i X$:

$$\begin{pmatrix} -X & A_i X + B_j \Gamma_i \\ (A_i X + B_j \Gamma_i)^T & -X \end{pmatrix} < 0$$
 (28)

$$\left[\alpha_{hl}X + \beta_{hl}\left(A_iX + B_j\Gamma_i\right) + \beta_{lh}\left(A_iX + B_j\Gamma_i\right)^T\right] \underset{h,l \in [1,m]}{<} 0 \tag{29}$$

with i, j = 1, ..., N, that can be solved using available software, e.g. the YALMIP toolbox [?] with SeDuMi solver [?].

Remark 1: In [?], it has been reported that, despite the idea of *poles*, as introduced, does not have a strict mathematical interpretation in the TS case, it has a strict connection with the dynamical behavior of the system, justifying, from the engineering point of view, the abuse of language. In fact, using LMI regions, it is possible to enforce transient performance specifications, such as decay rates and overshoot suppression (damping).

4. Fault tolerant control using Takagi-Sugeno virtual actuators

4.1. Takagi-Sugeno faulty model

In this work, two types of actuator faults are considered: multiplicative faults, i.e. changes in the effectiveness of the faulty actuators, and stuck faults, where the inputs delivered by the faulty actuators are blocked to constant values. In the first case, the generic subsystem in (14) becomes as follows:

$$x_i(k+1) = A_i x(k) + B_{f,i}(\phi(k)) u(k)$$
(30)

with:

$$B_{f,i}(\phi(k)) = B_i diag(\phi_1(k), \dots, \phi_{n_n}(k))$$
(31)

where B_i denotes the nominal input matrix, and $\phi_{\mu}(k) \in]0,1]$ represents the effectiveness of the μ -th actuator, such that the value $\phi_{\mu} = 1$ represents the healthy situation.

In the second case, the generic subsystem in (14) becomes:

$$x_i(k+1) = A_i x(k) + B_i^* u(k) + (B_i - B_i^*) \bar{u}(k)$$
(32)

where $\bar{u}(k)$ is the vector containing the values of the stuck control inputs and B_i^* is the matrix obtained from B_i by replacing the columns corresponding to the stuck actuators with zero vectors.

Then, in the case of multiplicative faults, the reference model subsystems in (19) are changed as follows:

$$x_{ref,i}(k+1) = A_i x_{ref}(k) + B_{f,i} \left(\hat{\phi}(k)\right) u_{ref}(k)$$
(33)

where $\hat{\phi}(k)$ is an estimation of the multiplicative actuator faults, while in the case of stuck faults, they become:

$$x_{ref,i}(k+1) = A_i x_{ref}(k) + B_i^* u_{ref}(k) + (B_i - B_i^*) \,\hat{\bar{u}}(k)$$
(34)

where $\hat{u}(k)$ is an estimation of the stuck actuator faults.

Then, under the assumption that $\hat{\phi}(k) \cong \phi(k)$ and $\hat{\bar{u}}(k) \cong \bar{u}(k)$ (the case where such an assumption does not hold, i.e. when there is uncertainty in the fault estimation, will be addressed by future research), the generic subsystem of the error model takes the form:

$$e_i(k+1) = A_i e(k) + B_{f,i} \left(\hat{\phi}(k)\right) \Delta u(k)$$
(35)

or:

$$e_i(k+1) = A_i e(k) + B_i^* \Delta u(k) \tag{36}$$

for multiplicative or stuck faults, respectively.

Remark 2: The change of the reference model subsystems will also require a modification of the values of the reference inputs $u_{ref}(k)$ used to generate the desired trajectory.

4.2. Takagi-Sugeno virtual actuator design

Here, the concept of virtual actuator introduced in [?] is extended to TS systems. The main idea of this FTC method is to reconfigure the faulty plant such that the nominal controller could still be used without need of retuning it. The plant with the faulty actuators is modified adding the virtual actuator block that masks the fault and allows the controller to see the same plant as before the fault.

In the case of multiplicative faults, the virtual actuator is static and can be expressed as:

$$\Delta u(k) = \left(\sum_{i=1}^{N} \rho_i \left(\vartheta(k)\right) B_{f,i} \left(\hat{\phi}(k)\right)\right)^{\dagger} \sum_{i=1}^{N} \rho_i \left(\vartheta(k)\right) B_i \Delta u_{c,i}(k)$$
(37)

where $\Delta u_{c,i}(k)$ are the outputs of the controller subsystems in (22), and the symbol † denotes the Moore-Penrose pseudoinverse. In this case, the fault-hiding property is achieved thanks to the fact that the product:

$$\left(\sum_{i=1}^{N} \rho_{i}\left(\vartheta(k)\right) B_{f,i}\left(\phi(k)\right)\right) \left(\sum_{i=1}^{N} \rho_{i}\left(\vartheta(k)\right) B_{f,i}\left(\phi(k)\right)\right)^{\dagger}$$
(38)

eliminates the effects of the multiplicative faults.

On the other hand, in the case of stuck faults, the fault tolerance is achieved using the reconfiguration structure expressed by:

$$\Delta u(k) = \sum_{i=1}^{N} \rho_i(\vartheta(k)) \, \Delta u_i(k) \tag{39}$$

with $\Delta u_i(k)$ given by:

$$\Delta u_i(k) = \Delta u_{c,i}(k) - M_{v,i} x_v(k) \tag{40}$$

where the virtual actuator state $x_v(k)$ is obtained through the inference:

$$x_v(k) = \sum_{i=1}^{N} \rho_i(\vartheta(k)) x_{v,i}(k)$$
(41)

with $x_{v,i}(k)$ provided by the TS virtual actuator subsystems, defined as follows:

IF
$$\vartheta_1(k)$$
 is M_{i1} AND ... AND $\vartheta_p(k)$ is M_{ip}
THEN $x_{v,i}(k+1) = (A_i + B_i^* M_{v,i}) x_{v,i}(k) + (B_i - B_i^*) \Delta u_{c,i}(k) \quad i = 1, ..., N$

$$(42)$$

being $M_{v,i}$ the virtual actuator subsystem gains.

Moreover, in order to achieve the fault-hiding property, the signal entering into the controller is slightly modified, such that the outputs of the controller subsystems in (22) become as follows:

$$\Delta u_{c,i}(k) = K_i \left(e(k) + x_v(k) \right) \tag{43}$$

When the stuck fault appears, the TS virtual actuator reconstructs the vector $\Delta u(k)$ from the outputs of the nominal controller subsystems $\Delta u_{c,i}(k)$, taking into account the fault occurrence. The faulty plant and the TS virtual actuator are called the reconfigured TS plant, which is connected to the nominal TS controller. If the reconfigured TS plant behaves like the nominal plant, the loop consisting of the reconfigured plant and the TS controller behaves like the nominal closed-loop system.

4.3. Reconfiguration analysis

In the following, it is shown that thanks to the introduction of the virtual actuator block, the augmented system can be brought to a block-triangular form.

Theorem 1. Consider the augmented system made up by the faulty error system (36), the reconfiguration structure (39)-(40), the virtual actuator (41)-(42) and the control law (43):

$$\begin{pmatrix}
e(k+1) \\
x_{v}(k+1)
\end{pmatrix} = \sum_{i=1}^{N} \rho_{i} (\vartheta(k)) \cdot \\
A_{i} + B_{i}^{*} K_{i} & B_{i}^{*} (K_{i} - M_{v,i}) \\
(B_{i} - B_{i}^{*}) K_{i} & A_{i} + B_{i}^{*} M_{v,i} + (B_{i} - B_{i}^{*}) K_{i}
\end{pmatrix} \begin{pmatrix}
e(k) \\
x_{v}(k)
\end{pmatrix}$$
(44)

Then, there exists a similarity transformation such that the state matrix of the augmented system in the new state variables is block-triangular, as follows:

$$A_{aug,i} = \begin{pmatrix} A_i + B_i K_i & 0\\ (B_i - B_i^*) K_i & A_i + B_i^* M_{v,i} \end{pmatrix}$$
(45)

Proof: The proof is straightforward, and comes from introducing the new state variable $x_1(k) \triangleq e(k) + x_v(k)$ and considering the state $\begin{pmatrix} x_1(k) & x_v(k) \end{pmatrix}^T$.

Looking at (45), it can be seen that the state $x_1(k)$ is affected by K_i through the matrix $A_i + B_i K_i$, while the state $x_v(k)$ is affected by $M_{v,i}$ through the matrix $A_i + B_i^* M_{v,i}$. Hence, the TS controller and the TS virtual actuator can be designed independently.

The design conditions presented in Section 3.3 can be applied to the case of virtual actuator design by making the changes $B_i \to B_i^*$ and $K_i \to M_{v,i}$.

5. Application to the PEM fuel cell case study

5.1. Reference model

Let us define the following reference model:

$$\dot{p}_{sm}^{ref} = \frac{\gamma R_a}{V_{sm}} \left\{ W_{cp}^{ref} \left[T_{atm} + \frac{T_{atm}}{\eta_{cp}} \left(\left(\frac{\hat{p}_{sm}}{p_{atm}} \right)^{\frac{\gamma - 1}{\gamma}} - 1 \right) \right] - k_{sm,out} \hat{T}_{sm} \left[p_{sm}^{ref} - w_{ca}^{ref} \frac{T_{st}}{V_{ca}} \right] \right\}$$
(46)

$$\dot{p}_{rm}^{ref} = \frac{R_a T_{rm}}{V_{rm}} k_{ca,out} \left(w_{ca}^{ref} \frac{T_{st}}{V_{ca}} - p_{rm}^{ref} \right) - \frac{R_a T_{rm}}{V_{rm}} k_{rm,out}^{ref} \left(\hat{p}_{rm} - p_{atm} \right)$$
(47)

$$\dot{w}_{ca}^{ref} = \chi_{O_{2},i} k_{sm,out} \left(p_{sm}^{ref} - w_{ca}^{ref} \frac{T_{st}}{V_{ca}} \right) R_{O_{2}}
- \hat{\chi}_{O_{2},o} k_{ca,out} \left(w_{ca}^{ref} \frac{T_{st}}{V_{ca}} - p_{rm}^{ref} \right) R_{O_{2}} - \frac{M_{O_{2}} n I_{st}}{4F}
+ (1 - \chi_{O_{2},i}) k_{sm,out} \left(p_{sm}^{ref} - w_{ca}^{ref} \frac{T_{st}}{V_{ca}} \right) R_{N_{2}}
- (1 - \hat{\chi}_{O_{2},o}) k_{ca,out} \left(w_{ca}^{ref} \frac{T_{st}}{V_{ca}} - p_{rm}^{ref} \right) R_{N_{2}}$$
(48)

where \hat{p}_{sm} , \hat{p}_{rm} , \hat{T}_{sm} and $\hat{\chi}_{O_2,o}$ are estimations of p_{sm} , p_{rm} , T_{sm} and $\chi_{O_2,o}$, respectively. The available measurements, denoted by p_{sm}^y , p_{rm}^y and p_{rm}^y , are filtered to deal with the noise, leading to the following filtered variables:

$$\hat{p}_{sm}(k) = \kappa_{p_{sm}} \hat{p}_{sm}(k-1) + (1 - \kappa_{p_{sm}}) p_{sm}^{y}(k)$$
(49)

$$\hat{p}_{rm}(k) = \kappa_{p_{rm}} \hat{p}_{rm}(k-1) + (1 - \kappa_{p_{rm}}) \, p_{rm}^{y}(k) \tag{50}$$

$$\hat{T}_{sm}(k) = \kappa_{T_{sm}} \hat{T}_{sm}(k-1) + (1 - \kappa_{T_{sm}}) T_{sm}^{y}(k)$$
(51)

with the filter parameters $\kappa_{p_{sm}}$, $\kappa_{p_{rm}}$ and $\kappa_{T_{sm}}$ chosen to be equal to 0.99.

On the other hand, $\hat{\chi}_{O_2,o}$ is obtained as (see (8)-(9)):

$$\hat{\chi}_{O_2,o} = \frac{\hat{y}_{O_2,o} M_{O_2}}{\hat{y}_{O_2,o} M_{O_2} + (1 - \hat{y}_{O_2,o}) M_{N_2}}$$
(52)

$$\hat{y}_{O_2,o} = \frac{\hat{m}_{O_2} R_{O_2}}{\hat{m}_{O_2} R_{O_2} + \hat{m}_{N_2} R_{N_2}} \tag{53}$$

where \hat{m}_{O_2} and \hat{m}_{N_2} are estimations of m_{O_2} and m_{N_2} , obtained from steadystate considerations about (5)-(6):

$$0 = \chi_{O_2,i} k_{sm,out} \left(\hat{p}_{sm} - \frac{\left(\hat{m}_{O_2} R_{O_2} + \hat{m}_{N_2} R_{N_2} \right) T_{st}}{V_{ca}} \right) - \hat{\chi}_{O_2,o} k_{ca,out} \left(\frac{\left(\hat{m}_{O_2} R_{O_2} + \hat{m}_{N_2} R_{N_2} \right) T_{st}}{V_{ca}} - \hat{p}_{rm} \right) - \frac{M_{O_2} n I_{st}}{4F}$$

$$(54)$$

$$0 = (1 - \chi_{O_2,i}) k_{sm,out} \left(\hat{p}_{sm} - \frac{(\hat{m}_{O_2} R_{O_2} + \hat{m}_{N_2} R_{N_2}) T_{st}}{V_{ca}} \right) - (1 - \hat{\chi}_{O_2,o}) k_{ca,out} \left(\frac{(\hat{m}_{O_2} R_{O_2} + \hat{m}_{N_2} R_{N_2}) T_{st}}{V_{ca}} - \hat{p}_{rm} \right)$$

$$(55)$$

Given \hat{p}_{sm} , \hat{p}_{rm} and I_{st} , (52)-(55) is a system of two nonlinear equations of variables \hat{m}_{O_2} and \hat{m}_{N_2} , that can be solved using some appropriate solver, e.g. fsolve in the Matlab Optimization Toolbox [?].

The reference model provides the state trajectory to be tracked by the real PEM fuel cell, starting from the reference inputs W_{cp}^{ref} and $k_{rm,out}^{ref}$. The values of the reference inputs to be fed to the reference model (feedforward actions) are obtained from steady-state considerations about the fuel cell system, so as to keep the supply manifold pressure and the oxygen stoichiometry at some desired values p_{sm}^{∞} and $\lambda_{O_2}^{ref}$. In particular, the following nonlinear equations are obtained:

$$W_{cp}^{ref} \left[T_{atm} + \frac{T_{atm}}{\eta_{cp}} \left[\left(\frac{\hat{p}_{sm}}{p_{atm}} \right)^{\frac{\gamma - 1}{\gamma}} - 1 \right] \right] -k_{sm,out} \hat{T}_{sm} \left(p_{sm}^{\infty} - w_{ca}^{\infty} \frac{T_{st}}{V_{ca}} \right) T_{sm} = 0$$

$$(56)$$

$$k_{ca,out}w_{ca}^{\infty} \left(\frac{T_{st}}{V_{ca}} - p_{rm}^{\infty}\right) - k_{rm,out}^{ref} \left(\hat{p}_{rm} - p_{atm}\right) = 0$$

$$(57)$$

$$\chi_{O_{2},i}k_{sm,out} \left(p_{sm}^{\infty} - w_{ca}^{\infty} \frac{T_{st}}{V_{ca}} \right) R_{O_{2}} \\
-\hat{\chi}_{O_{2},o}k_{ca,out} \left(w_{ca}^{\infty} \frac{T_{st}}{V_{ca}} - p_{rm}^{\infty} \right) R_{O_{2}} - \frac{M_{O_{2}}nI_{st}}{4F} \\
+ (1 - \chi_{O_{2},i}) k_{sm,out} \left(p_{sm}^{\infty} - w_{ca}^{\infty} \frac{T_{st}}{V_{ca}} \right) R_{N_{2}} \\
- (1 - \hat{\chi}_{O_{2},o}) k_{ca,out} \left(w_{ca}^{\infty} \frac{T_{st}}{V_{ca}} - p_{rm}^{\infty} \right) R_{N_{2}} = 0$$
(58)

$$\lambda_{O_2}^{ref} - \frac{\chi_{O_2,i} k_{sm,out} \left(p_{sm}^{\infty} - w_{ca}^{\infty} \frac{T_{st}}{V_{ca}} \right)}{M_{O_2} \frac{nI_{st}}{4F}} = 0$$
 (59)

Given \hat{p}_{sm} , \hat{p}_{rm} , \hat{T}_{sm} , $\hat{\chi}_{O_2,o}$ and I_{st} , and defined the desired values p_{sm}^{∞} and $\lambda_{O_2}^{ref}$, (56)-(59) is a system of four nonlinear equations of variables W_{cp}^{ref} , $k_{rm,out}^{ref},\,w_{ca}^{\infty}$ and $p_{rm}^{\infty},$ that can be solved using fsolve [?].

5.2. Error model

By subtracting the reference model equations (46)-(48) and the corresponding system equations (2), (4) and (11), and by defining the tracking errors $e_1 \triangleq p_{sm}^{ref} - p_{sm}$, $e_2 \triangleq p_{rm}^{ref} - p_{rm}$, $e_3 \triangleq w_{ca}^{ref} - w_{ca}$, and the new inputs $\Delta u_1 \triangleq W_{cp}^{ref} - W_{cp}$, $\Delta u_2 \triangleq k_{rm,out}^{ref} - k_{rm,out}$, by considering $\hat{T}_{sm} \cong T_{sm}$, and $\hat{\chi}_{O_2,o} \cong \chi_{O_2,o}$, the error model for the PEM fuel cell can be brought to the following representation:

$$\dot{e}_{1} = -\frac{\gamma R_{a}}{V_{sm}} k_{sm,out} \hat{T}_{sm} \left(e_{1} - \frac{T_{st}}{V_{ca}} e_{3} \right) + b_{11} \left(\hat{p}_{sm} \right) \Delta u_{1}$$
 (60)

$$\dot{e}_{2} = -\frac{R_{a}T_{rm}}{V_{rm}}k_{ca,out}\left(e_{2} - \frac{T_{st}}{V_{ca}}e_{3}\right) + b_{22}\left(\hat{p}_{rm}\right)\Delta u_{2} \tag{61}$$

$$\dot{e}_{3} = k_{sm,out} \left[\chi_{O_{2},i} R_{O_{2}} + (1 - \chi_{O_{2},i}) R_{N_{2}} \right] \left(e_{1} - \frac{T_{st}}{V_{ca}} e_{3} \right)$$

$$+ k_{ca,out} \left[\chi_{O_{2},o} R_{O_{2}} + (1 - \chi_{O_{2},o}) R_{N_{2}} \right] \left(e_{2} - \frac{T_{st}}{V_{ca}} e_{3} \right)$$

$$(62)$$

with:

$$b_{11}\left(\hat{p}_{sm}\right) = \frac{\gamma R_a}{V_{sm}} \left[T_{atm} + \frac{T_{atm}}{\eta_{cp}} \left[\left(\frac{\hat{p}_{sm}}{p_{atm}} \right)^{\frac{\gamma - 1}{\gamma}} - 1 \right] \right]$$
(63)

$$b_{22}(\hat{p}_{rm}) = -\frac{R_a T_{rm}}{V_{rm}} (\hat{p}_{rm} - p_{atm})$$
(64)

5.3. Fault definition

In this work, three possible faults are considered.

5.3.1. Fault 1

Loss of effectiveness of the compressor W_{cp}

By including this fault in the dynamic model of the PEM fuel cell, (2) becomes:

$$\dot{p}_{sm} = \frac{\gamma R_a}{V_{sm}} \phi_{cp} W_{cp} \left[T_{atm} + \frac{T_{atm}}{\eta_{cp}} \left[\left(\frac{p_{sm}}{p_{atm}} \right)^{\frac{\gamma - 1}{\gamma}} - 1 \right] \right] - \frac{\gamma R_a}{V_{sm}} k_{sm,out} T_{sm} \left(p_{sm} - w_{ca} \frac{T_{st}}{V_{ca}} \right)$$

$$(65)$$

where ϕ_{cp} denotes the multiplicative fault of the air compressor flow.

5.3.2. Fault 2

Loss of effectiveness of the return manifold valve $k_{rm,out}$ By including this fault, (4) changes to:

$$\dot{p}_{rm} = \frac{R_a T_{rm}}{V_{rm}} k_{ca,out} \left(w_{ca} \frac{T_{st}}{V_{ca}} - p_{rm} \right) - \frac{R_a T_{rm}}{V_{rm}} \phi_{rm} k_{rm,out} \left(p_{rm} - p_{atm} \right)$$
(66)

where ϕ_{rm} denotes the multiplicative fault of the return manifold outlet orifice.

5.3.3. Fault 3

Stuck of the return manifold valve $k_{rm,out}$ In this case, $k_{rm,out}$ is stuck to $\bar{k}_{rm,out}$, such that (4) becomes:

$$\dot{p}_{rm} = \frac{R_a T_{rm}}{V_{rm}} k_{ca,out} \left(w_{ca} \frac{T_{st}}{V_{ca}} - p_{rm} \right) - \frac{R_a T_{rm}}{V_{com}} \bar{k}_{rm,out} \left(p_{rm} - p_{atm} \right)$$

$$(67)$$

5.4. Fault detection/estimation

In order to apply the proposed strategy, a fault estimation is needed. Hereafter, the fault estimation is formulated as a parameter estimation problem in such a way that any parameter estimation algorithm, such as least squares, could be used. In general, least squares (LS) algorithms can be formulated either in block or in recursive online form [?]. Once the equation is put in regressor form, the recursive formulation [?] and the block formulation [?] are interchangeable.

For estimating the loss of effectiveness of the compressor ϕ_{cp} , the discretetime version of (65), obtained using an Euler approach with sampling time T_s , is considered and put in the following regressor form:

$$z_{sm}(k) = \mu_{sm}(k-1)\phi_{cp}(k-1)$$
(68)

with:

$$z_{sm}(k) = p_{sm}(k) - p_{sm}(k-1) + \frac{\gamma R_a}{V_{sm}} k_{sm,out} T_{sm}(k-1) \left(p_{sm}(k-1) - w_{ca}(k-1) \frac{T_{st}}{V_{ca}} \right) T_s$$
(69)

$$\mu_{sm}(k) = \frac{\gamma R_a}{V_{sm}} W_{cp}(k) \left\{ T_{atm} + \frac{T_{atm}}{\eta_{cp}} \left[\left(\frac{p_{sm}(k)}{p_{atm}} \right)^{\frac{\gamma - 1}{\gamma}} - 1 \right] \right\} T_s$$
 (70)

Similarly, for detecting the presence of a fault in the return manifold valve, the following regressor form is used, that will allow to detect the presence of either a loss of effectiveness fault as in (66) or a stuck fault¹ as in (67):

$$z_{rm}(k) = \mu_{rm}(k-1)\phi_{rm,out}(k-1)$$
(71)

with:

$$z_{rm}(k) = p_{rm}(k) - p_{rm}(k-1) - \frac{R_a T_{rm}}{V_{rm}} k_{ca,out} \left(w_{ca}(k-1) \frac{T_{st}}{V_{ca}} - p_{rm}(k-1) \right) T_s$$
(72)

¹Since the effect of the stuck fault will lead to a difference between the real value of the return manifold valve outlet constant, and the control input, a change in ϕ_{rm} will appear not only in the case of a loss of effectiveness fault, but in the case of a stuck fault too.

$$\mu_{rm}(k) = -\frac{R_a T_{rm}}{V_{rm}} k_{rm,out}(k) \left(p_{rm}(k) - p_{atm} \right) T_s$$
 (73)

Then, if a block formulation with time window N_{LS} is used, the LS fault estimations are obtained as:

$$\hat{\phi}_{sm}(k) = M_{sm}(k)^{\dagger} \xi_{sm}(k) \tag{74}$$

$$\hat{\phi}_{rm}(k) = M_{rm}(k)^{\dagger} \xi_{rm}(k) \tag{75}$$

$$\xi_{sm}(k) = \begin{bmatrix} z_{sm}(k) \\ z_{sm}(k-1) \\ \vdots \\ z_{sm}(k-N_{LS}+1) \end{bmatrix} M_{sm}(k) = \begin{bmatrix} \mu_{sm}(k-1) \\ \mu_{sm}(k-2) \\ \vdots \\ \mu_{sm}(k-N_{LS}) \end{bmatrix}$$

$$\xi_{rm}(k) = \begin{bmatrix} z_{rm}(k) \\ z_{rm}(k-1) \\ \vdots \\ z_{rm}(k-N_{LS}+1) \end{bmatrix} M_{rm}(k) = \begin{bmatrix} \mu_{rm}(k-1) \\ \mu_{rm}(k-2) \\ \vdots \\ \mu_{rm}(k-N_{LS}) \end{bmatrix}$$

with $M_{sm}(k)^{\dagger}$ and $M_{rm}(k)^{\dagger}$ denoting the pseudoinverses of $M_{sm}(k)$ and $M_{rm}(k)$, respectively.

5.5. Faulty error model of the PEM Fuel Cell

In the case of loss of effectiveness due to compressor fault, the reference model equation (46) is modified as follows:

$$\dot{p}_{sm}^{ref} = \frac{\gamma R_a}{V_{sm}} \left\{ \hat{\phi}_{cp} W_{cp}^{ref} \left[T_{atm} + \frac{T_{atm}}{\eta_{cp}} \left(\left(\frac{\hat{p}_{sm}}{p_{atm}} \right)^{\frac{\gamma - 1}{\gamma}} - 1 \right) \right] - k_{sm,out} \hat{T}_{sm} \left[p_{sm}^{ref} - w_{ca}^{ref} \frac{T_{st}}{V_{ca}} \right] \right\}$$
(76)

Then, under the assumption that $\hat{\phi}_{cp} \cong \phi_{cp}$, the error equation (60) becomes:

$$\dot{e}_{1} = -\frac{\gamma R_{a}}{V_{sm}} k_{sm,out} \hat{T}_{sm} \left(e_{1} - \frac{T_{st}}{V_{ca}} e_{3} \right) + b_{11} \left(\hat{p}_{sm} \right) \hat{\phi}_{cp} \Delta u_{1}$$
 (77)

Notice that a slight modification of (56) is needed, i.e. replacing W_{cp}^{ref} with $\hat{\phi}_{cp}W_{cp}^{ref}$.

On the other hand, when a fault in the return manifold valve is detected, an estimation of the real value of $k_{rm,out}$, denoted in the following as $\hat{k}_{rm,out}$, is obtained using a LS approach similar to the one described in Section 5.4, using the following regressor equation:

$$z_{rm}(k) = \mu_{krm}(k-1)k_{rm,out}(k-1)$$
(78)

with $z_{rm}(k)$ defined as in (72) and $\mu_{krm}(k)$ defined as:

$$\mu_{krm}(k) = -\frac{R_a T_{rm}}{V_{rm}} \left(p_{rm}(k) - p_{atm} \right) T_s \tag{79}$$

Then, the reference model equation (47) is modified as follows:

$$\dot{p}_{rm}^{ref} = \frac{R_a T_{rm}}{V_{rm}} k_{ca,out} \left(w_{ca}^{ref} \frac{T_{st}}{V_{ca}} - p_{rm}^{ref} \right) - \frac{R_a T_{rm}}{V_{rm}} \hat{k}_{rm,out} \left(\hat{p}_{rm} - p_{atm} \right)$$
(80)

and the control input $k_{rm,out}(k)$ is chosen as $\hat{k}_{rm,out}/\hat{\phi}_{rm}$. Then, under the assumption that $\hat{\phi}_{rm} \cong \phi_{rm}$ in the loss of effectiveness fault case, and $\hat{k}_{rm,out} \cong \bar{k}_{rm,out}$ in the stuck fault case, in both cases the error equation (61) becomes:

$$\dot{e}_2 = -\frac{R_a T_{rm}}{V_{rm}} k_{ca,out} \left(e_2 - \frac{T_{st}}{V_{ca}} e_3 \right) \tag{81}$$

Also, (57) should be modified by replacing $k_{rm,out}^{ref}$ with $\hat{k}_{rm,out}$. Then, the degree of freedom in the choice of p_{sm}^{∞} is lost, and (56)-(59) become a system of four nonlinear equations of variables W_{cp}^{ref} , p_{sm}^{∞} , w_{ca}^{∞} and p_{rm}^{∞} .

6. Simulation results

By considering interval bounds for p_{sm} and p_{rm} , i.e. $p_{sm} \in [1.3 \, Pa, 2.3 \, Pa]$ and $p_{rm} \in [1.3 \, Pa, 2.3 \, Pa]$, minimum and maximum values for $b_{11}(p_{sm})$ and $b_{22}(p_{rm})$ are obtained, as follows: $b_{11} \in [6.54 \cdot 10^6, 7.96 \cdot 10^6]$, $b_{22} \in [-2.21 \cdot 10^{12}, -0.491 \cdot 10^{12}]$. Then, by taking into account the bounds for T_{sm} and $\chi_{O_2,o}$, i.e. $T_{sm} \in [250 \, K, 400 \, K]$ and $\chi_{O_2,o} \in [0.07, 0.21]$, it is possible to use the sector nonlinearity approach [?] for the varying elements of the state matrix obtained from (60)-(62), the gridding approach for $b_{11}(p_{sm})$ and $b_{22}(p_{rm})$ with $N_{11} = 12$ and $N_{22} = 12$ points, respectively, and an Euler discretization with sampling time $T_s = 0.01 \, s$ for transforming (60)-(62) into an equivalent TS model, as follows:

$$IF \vartheta_1(k) is M_{i1} AND \vartheta_2(k) is M_{i2} \cdots AND \vartheta_4(k) is M_{i4}$$

$$THEN e_i(k+1) = A_i e(k) + B_i \Delta u(k) \quad i = 1, \dots, N$$
(82)

where $\vartheta_1(k) = \hat{T}_{sm}(k)$, $\vartheta_2(k) = \hat{\chi}_{O_2,o}(k)$, $\vartheta_3(k) = b_{11}(p_{sm}(k))$ and $\vartheta_4(k) = b_{22}(p_{rm}(k))$ are the premise variables, and $N = 4N_{11}N_{22} = 576$.

On the other hand, when considering the loss of effectiveness of the compressor (fault 1), the error model becomes:

IF
$$\vartheta_1(k)$$
 is M_{i1} AND $\vartheta_2(k)$ is $M_{i2} \cdots AND \ \vartheta_4(k)$ is M_{i4}
THEN $e_i(k+1) = A_i e(k) + B_{f,i}(\hat{\phi}_{cp}) \Delta u(k) \ i = 1, \dots, N$ (83)

Finally, in the cases of loss of effectiveness of return manifold valve or stuck

of return manifold valve (faults 2 and 3, respectively), the error model becomes:

IF
$$\vartheta_1(k)$$
 is M_{i1} AND $\vartheta_2(k)$ is $M_{i2} \cdots$ AND $\vartheta_4(k)$ is M_{i4}
THEN $e_i(k+1) = A_i e(k) + B_i^* \Delta u(k) \quad i = 1, \dots, N$ (84)

where the matrices B_i^* are obtained from B_i by replacing the second column (the one corresponding to the input $k_{rm,out}$) with a zero column.

When using a gridding approach to obtain a TS model, as is the case of premise variables $b_{11}(p_{sm})$ and $b_{22}(p_{sm})$, it is possible to reduce the overall number of LMIs by requiring the stability and desired performance to be strictly guaranteed only at the design points. In this case, as long as the gridding is sufficiently dense, it is reasonable to assume that stability and performance will still hold at operating points different from the design ones. In the case of the PEM fuel cell, this allows to reduce the total number of LMIs to be solved from 331777 to 577.

In particular, the nominal controller has been designed to assure stability and pole clustering in a circle of radius 0.4 and center (0.4,0). The dynamic virtual actuator that takes into account the faults 2 and 3 has been designed using a circle of radius 0.3 and center (0.4,0).

The results shown in this paper refer to simulations that last 200 s, where the current in the stack I_{st} (load) and the desired oxygen stoichiometry vary in time, as follows:

$$I_{st}(t) = \begin{cases} 100 & t \le 40s \\ 70t - 2700 & 40s < t \le 41s \\ 170 & 41s < t \le 80s \\ -50t + 4170 & 80s < t \le 81s \\ 120 & 81s < t \le 120s \\ 40t - 4680 & 120s < t \le 121s \\ 160 & 121s < t \le 160s \\ -10t + 1760 & 160s < t \le 161s \\ 150 & 161s < t \le 200s \end{cases}$$
(85)

$$\lambda_{O_2}^{ref}(t) = \begin{cases}
2 & t \le 40s \\
-0.3t + 14 & 40s < t \le 41s \\
1.7 & 41s < t \le 80s \\
0.4t - 30.3 & 80s < t \le 81s \\
2.1 & 81s < t \le 120s \\
-0.2t + 26.1 & 120s < t \le 121s \\
1.9 & 121s < t \le 160s \\
0.1t - 14.1 & 160s < t \le 161s \\
2 & 161s < t \le 200s
\end{cases} \tag{86}$$

and the desired supply manifold pressure is set to $p_{sm}^{\infty} = 1.5 \, Pa$. In addition, the simulations have been performed assuming uncertainty in some parameters, as resumed in Table 1. The nominal values have been taken from [?], and the uncertainty has been generated for each parameter by considering a uniform

Table 1: Nominal and real parameter values.

Parameter	Nominal value	Real value
γ	1.4	=
η_{cp}	0.8	0.799
$\chi_{O_2,i}$	0.233	=
$k_{ca,out}$	$2.177 \cdot 10^{-6}$	$2.170 \cdot 10^{-6}$
$k_{sm,out}$	$3.629 \cdot 10^{-6}$	$3.601 \cdot 10^{-6}$
n	381	=
p_{atm}	101325	101040
$y_{O_2,i}$	0.21	=
F	96485	=
M_{N_2}	$28 \cdot 10^{-3}$	=
M_{O_2}	$32 \cdot 10^{-3}$	=
R_a	286.9	=
R_{N_2}	296.8	=
R_{O_2}	259.8	=
T_{atm}	298.15	296.24
T_{rm}	300	298.69
T_{st}	350	348.87
V_{ca}	0.01	0.0099
V_{rm}	0.005	=
V_{sm}	0.002	=

aleatory distribution around 1% of the nominal value. The noise in the sensor measurements has been considered as uniformly distributed around 1% of the measurement. Finally, the dynamic virtual actuator is activated when $\hat{\phi}_{rm} < 0.8$.

Remark 2: The choice of the oxygen stoichiometry $\lambda_{O_2}^{ref}$ and the current in the stack I_{st} is inspired by [?], where the control system is tested against their changes, due to different required net powers. As remarked in several works, e.g. [??], the oxygen excess ratio should be regulated around a value of 2 to reduce the formation of stagnant vapor and nitrogen films in the electrochemical area, and prevent the oxygen starvation phenomenon, which has serious consequences on the stack life. In this work, following [?], a non-constant $\lambda_{O_2}^{ref}$ is considered in order to show the capability of the proposed approach to deal with operating point changes. It is also worth noticing that abrupt changes of $I_{st}(t)$ could be considered as well, even though they would lead to an increase of the oscillatory transients, similarly to the results shown in [?].

6.1. Fault scenario 1

In fault scenario 1, a loss of effectiveness $\phi_{cp}=0.5$ appearing at time $t=100\,s$ has been considered. Fig. 1 shows a comparison of the oxygen stoichiometry λ_{O_2} obtained in both the cases when the nominal controller is used without the proposed FTC strategy (red line) and when the proposed FTC strategy is applied (blue line). The improvement in the tracking performance

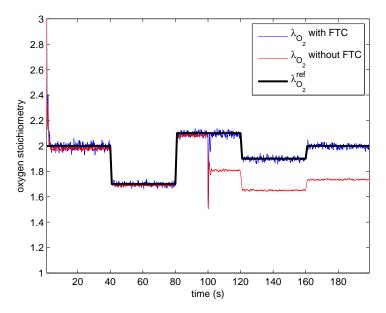


Figure 1: Oxygen stoichiometry λ_{O_2} in fault scenario 1.

brought by the virtual actuator can be seen clearly. Fig. 2 shows the estimation of ϕ_{cp} , demonstrating that the least-squares based algorithm proposed in Section 5.4 is able to estimate correctly the fault (a steady-state offset appears due to the uncertainty in the model used for the simulations). The virtual actuator reconfigures the control input (see blue line in Fig. 3) such that the effect of the control input taking into account the fault (see cyan line in Fig. 3) matches the control input that would be obtained in nominal conditions (see black line in Fig. 3). On the other hand, when no FTC strategy is applied, the intrinsic robustness of the nominal controller against faults due to the feedback action (see red line in Fig. 3) is not sufficient to compensate the fault effect, since there appears a mismatch with respect to the nominal case (see magenta line in Fig. 3). Finally, Fig. 4 shows the estimation of $\chi_{O_2,o}$ obtained using the methodology described in Section 5.1. Also in this case, the algorithm succeeds in estimating correctly the oxygen mass fraction in the outlet.

6.2. Fault scenario 2

In fault scenario 2, a loss of effectiveness $\phi_{rm} = 0.5$ appearing at time $t = 100 \, s$ has been considered. Similarly to the case of fault scenario 1, an offset appears in the response of λ_{O_2} when no FTC strategy is applied (red line in Fig. 5). Despite not being able to reduce the offset to zero, due to the presence of estimation errors caused by the uncertainty, the proposed FTC strategy is able to reduce strongly the effect of the fault on the tracking performance (see blue line in Fig. 5). The dynamic virtual actuator, activated when the estimation of

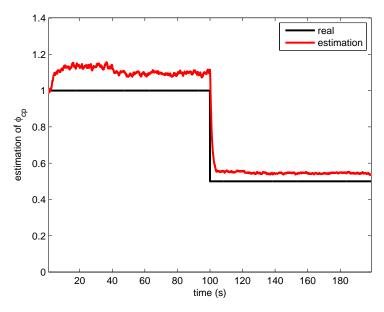


Figure 2: Estimation of ϕ_{cp} in fault scenario 1.

 ϕ_{rm} goes below 0.8 (see Fig. 6) achieves the fault tolerance by redistributing the controller output on the remaining healthy actuator (the compressor), as shown in Fig. 7.

6.3. Fault scenario 3

In fault scenario 3, the return manifold outlet is stuck starting from time $t=100\,s$. As shown in Fig. 8, no effect is visible in the oxygen excess ratio response until a change in the reference $\lambda_{O_2}^{ref}$ occurs at time $t=120\,s$. This fact is reasonable, because in the interval between the fault occurrence and the reference change, the return manifold outlet is stuck near the correct position that assures an error approximately zero. After the reference change, it can be seen that an offset appears due to the fault (red line in Fig. 8). Also in this case, the reconfiguration brought by the dynamic virtual actuator improves the tracking performance under fault occurrence (blue line in Fig. 8).

7. Conclusions

In this paper, an FTC strategy based on the use of virtual actuators in the TS framework for PEM fuel cells has been proposed. The overall solution relies on adding a virtual actuator in the control loop to hide the fault from the controller point of view, allowing it to see the same plant as before the fault, in this way keeping the stability and some desired performances. The proposed

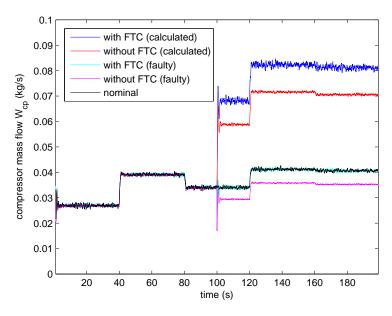


Figure 3: Compressor mass flow W_{cp} in fault scenario 1.

methodology is based on the use of a reference model, where the resulting non-linear error model is brought to a TS form using the gridding approach. The TS model is suitable for designing a controller using LMI-based techniques, such that the resulting closed-loop error system is stable with poles placed in some desired region of the complex plane. Simulation results have shown that if no FTC strategy is applied, undesired offsets would appear. On the other hand, the proposed FTC strategy allows to improve the overall performance in all the considered cases. As future work, the proposed approach will be extended to the case of sensor faults using the virtual sensor technique.

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Nomenclature

- γ denotes the specific heat capacity of gas
- η_{cp} denotes the compressor efficiency
- $\chi_{O_2,i}$ denotes the oxygen mass fraction in the inlet

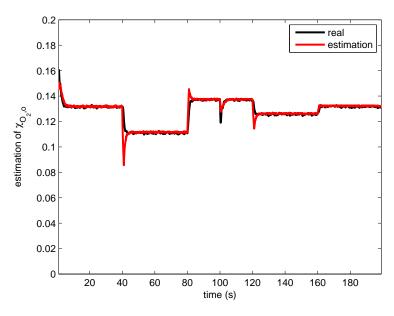


Figure 4: Estimation of $\chi_{O_2,o}$ in fault scenario 1.

- $\chi_{O_2,o}$ denotes the oxygen mass fraction in the outlet
- $k_{ca,out}$ denotes the cathode outlet flow constant (ms)
- $k_{rm,out}$ denotes the return manifold outlet flow constant (ms)
- $k_{sm.out}$ denotes the supply manifold outlet flow constant (ms)
- m_{N_2} denotes the mass of nitrogen in the cathode (kg)
- m_{O_2} denotes the mass of oxygen in the cathode (kg)
- m_{sm} denotes the mass of gas in the supply manifold (kg)
- n denotes the number of cells in the FCS
- p_{atm} denotes the air pressure (Pa)
- p_{rm} denotes the return manifold pressure (Pa)
- p_{sm} denotes the supply manifold pressure (Pa)
- $y_{O_2,i}$ denotes the oxygen mole fraction in the inlet
- $y_{O_2,o}$ denotes the oxygen mole fraction in the outlet
- F denotes the Faraday constant (C/mol)
- I_{st} denotes the current in the stack (A)
- M_{N_2} denotes the nitrogen molar mass (kg/mol)
- M_{O_2} denotes the oxygen molar mass (kg/mol)
- R_a denotes the air gas constant (J/(kgK))
- R_{N_2} denotes the nitrogen gas constant (J/(kgK))
- $R_{O_{2}}$ denotes the oxygen gas constant $\left(J/\left(kgK\right) \right)$
- T_{atm} denotes the air temperature (K)
- T_{rm} denotes the return manifold temperature (K)
- T_{sm} denotes the supply manifold temperature (K)

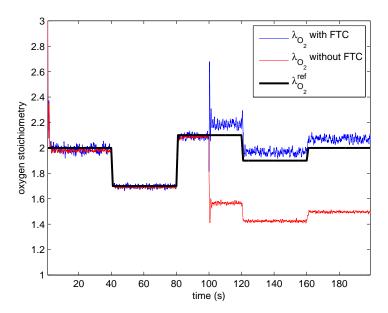


Figure 5: Oxygen stoichiometry λ_{O_2} in fault scenario 2.

- T_{st} denotes the stack temperature (K) V_{ca} denotes the cathode volume (m^3)
- V_{rm} denotes the return manifold volume (m^3) V_{sm} denotes the supply manifold volume (m^3)
- W_{cp} denotes the compressor air mass flow rate (kg/s)

References

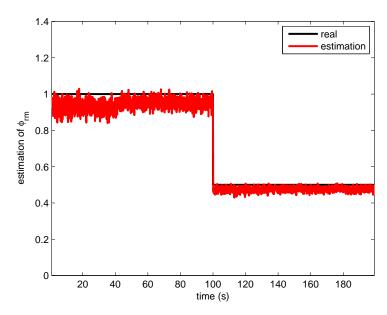


Figure 6: Estimation of ϕ_{rm} in fault scenario 2.

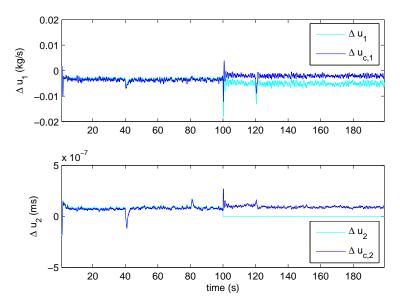


Figure 7: Controller output Δu_c and reconfiguration brought by the dynamic virtual actuator Δu in fault scenario 2.

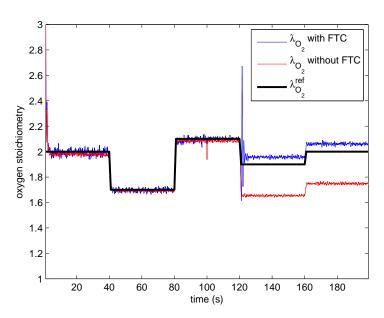


Figure 8: Oxygen stoichiometry λ_{O_2} in fault scenario 3.